

SEE 3243

Registers & Counters

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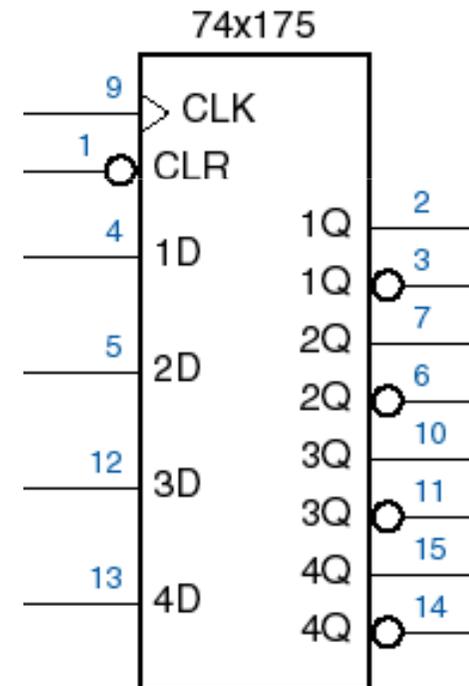
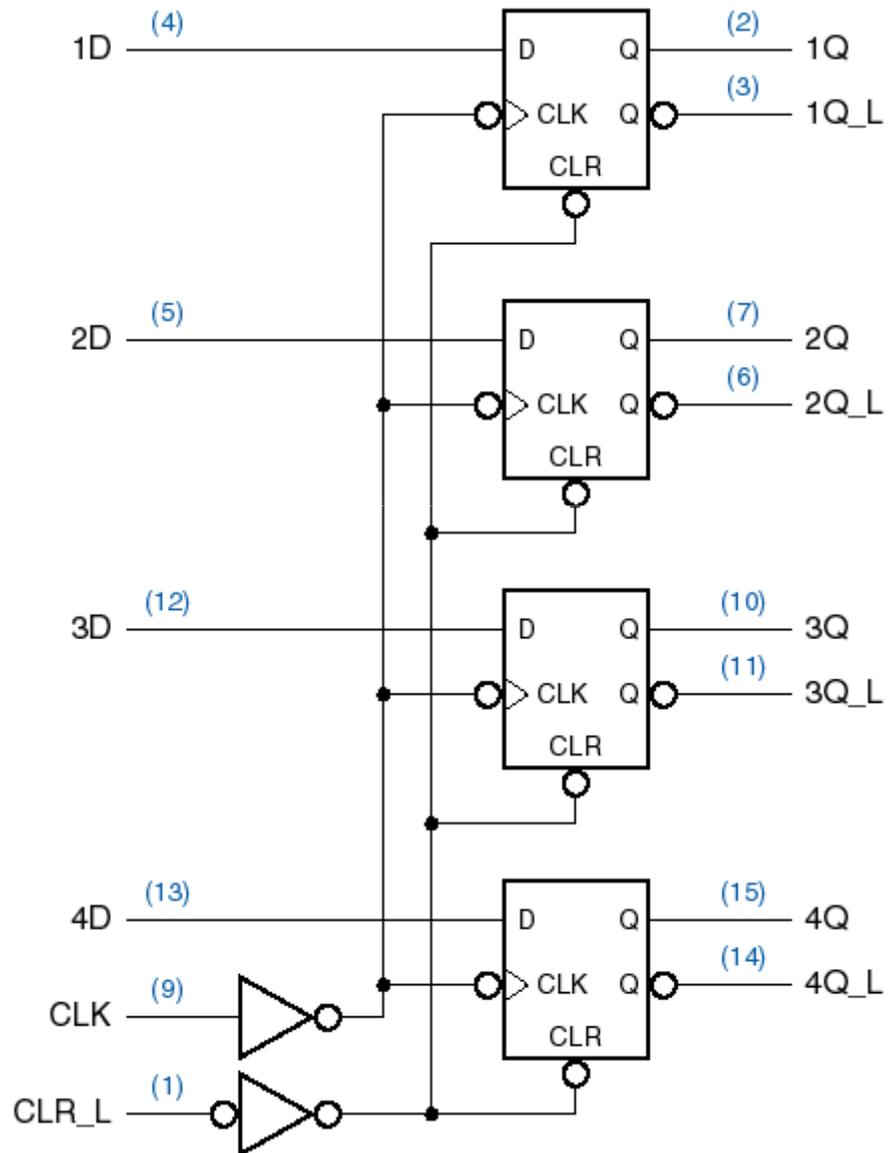
Week 8

- Storage Registers
- Shift Registers
- Counters
- Design of Synchronous Counters

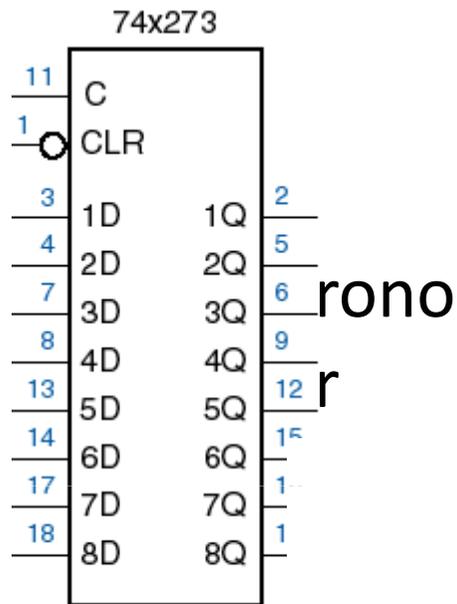
Registers

- Register is a group of flip-flops/memory elements that work together to store data or instructions and shift a group of bits or a binary word.
- Variations:
 - Register file - a few registers, each accessible by a register address. Sort of a small memory array.
 - Shift register - temporary circuit able to shift or move the stored word either left or right.
 - the bits stored can be moved/shifted from 1 element to another adjacent element.
 - all the storage registers are actuated simultaneously by a single input clock/shift pulse.
 - Buffer register - a temporary data storage circuit able to store a digital word.
- Sometimes use special names
 - accumulators, program counters, index registers, stack pointer, status register, etc.

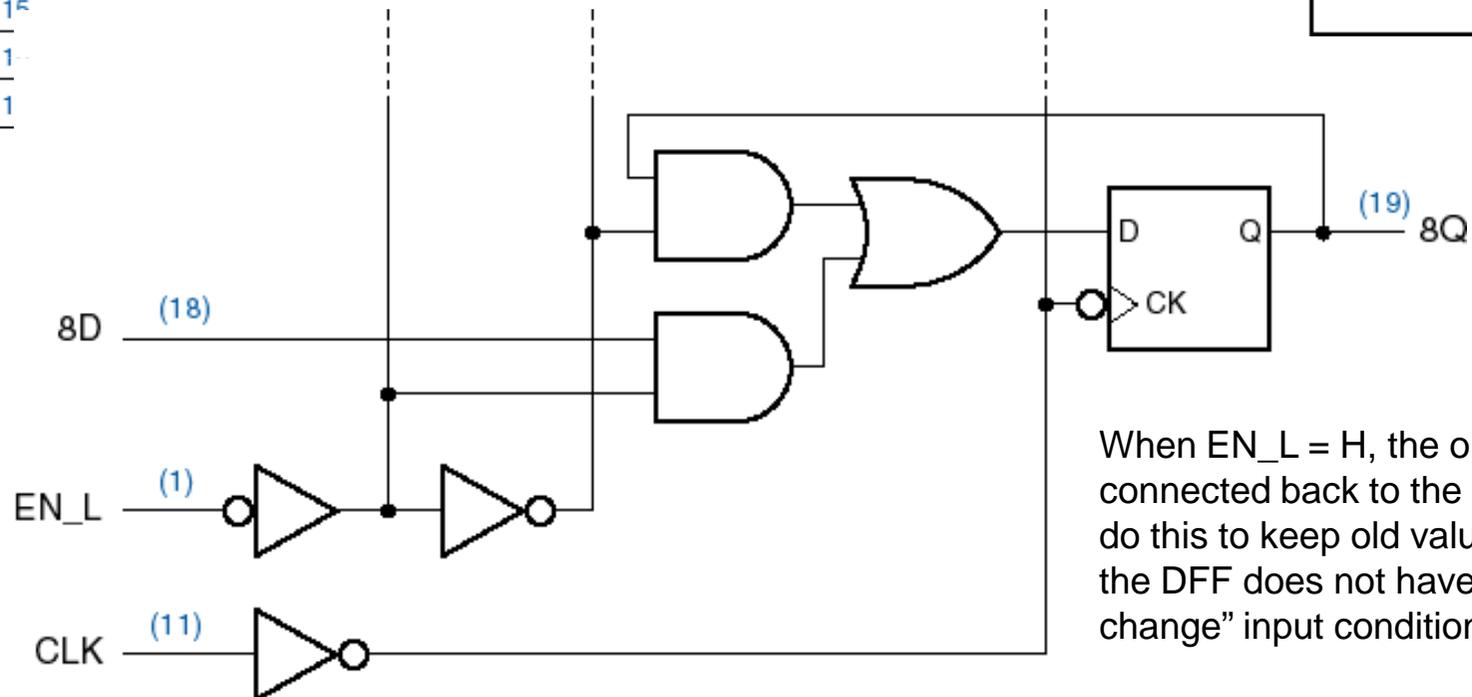
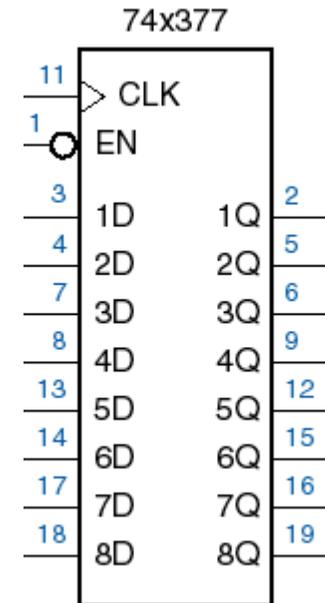
Multibit registers and latches



Octal (8-bit) Register & Latch



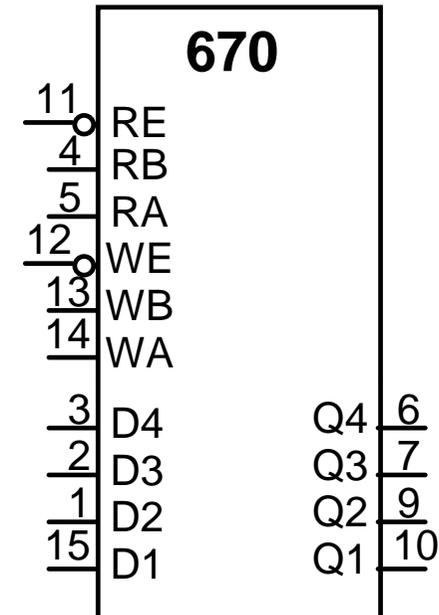
- 74x377
- clock enable



When EN_L = H, the output is connected back to the input. Must do this to keep old value because the DFF does not have a "no change" input condition

74x670 4x4 Register File with Tri-state Outputs

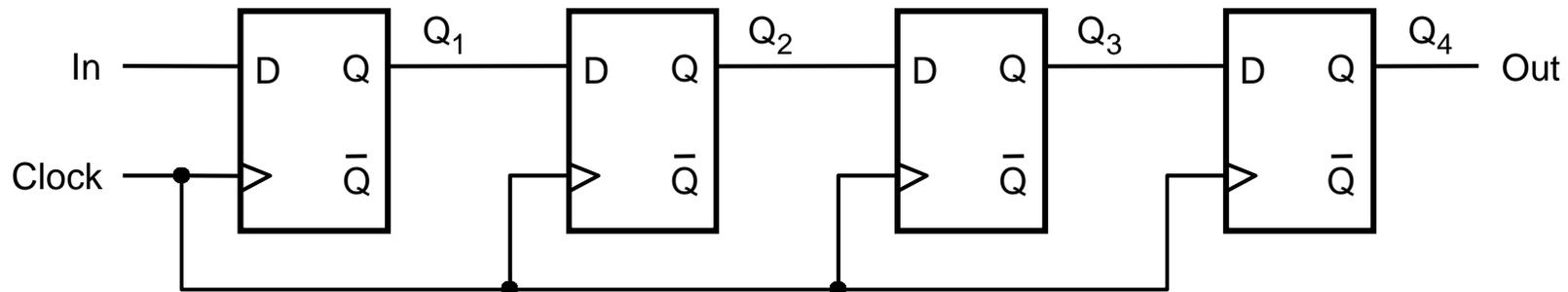
- The 74x670 device contains 16 D flip-flops organized into four words of four flip-flops each.
- Each register in the register file is called a **word** and is identified by a unique index or **address**
- Word contents read or written
 - Separate Read and Write Enables (RE, WE)
 - Separate Read and Write Address (RA, RB, WA, WB) - binary encodings of one of four registers to be read or written
 - Data Input, Q Outputs
- On a read, the selected word is multiplexed to the outputs.
- On a write, data present on D4-D1 inputs are stored in the selected word



Shift Registers

- Register components that shift as well as store
- For handling serial data, such as RS-232 and modem transmission and reception, Ethernet links, SONET, etc.
- Data moves from left to right (or from top to bottom). On every shift pulse, the contents of a given flip-flop are replaced by the contents of the flip-flop to its left. The leftmost device receives its inputs from the rightmost.
- Because flip-flop propagation times far exceed hold times, the values are passed correctly from one stage to the next

Basic Shift Register



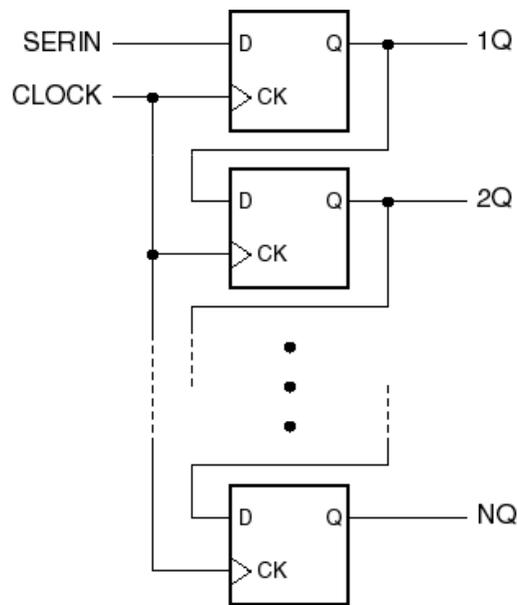
(a) Circuit

	In	Q ₁	Q ₂	Q ₃	Q ₄ = Out
t_0	1	0	0	0	0
t_1	0	1	0	0	0
t_2	1	0	1	0	0
t_3	1	1	0	1	0
t_4	1	1	1	0	1
t_5	0	1	1	1	0
t_6	0	0	1	1	1
t_7	0	0	0	1	1

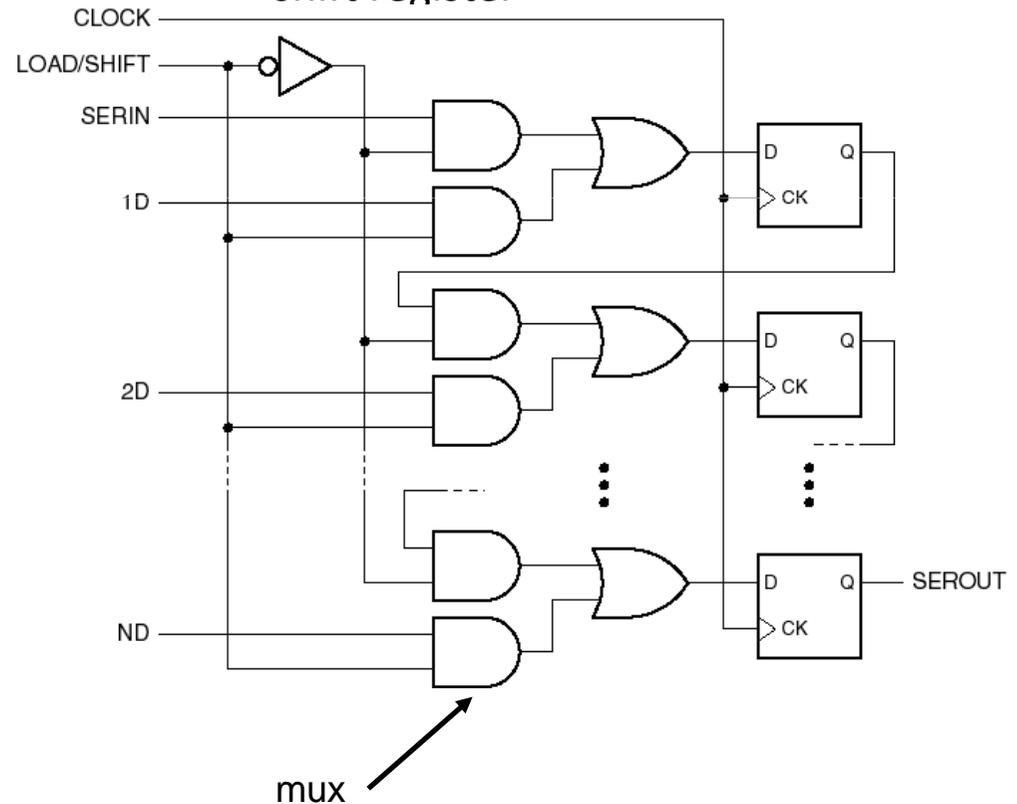
(b) A sample sequence

Parallel-to-serial conversion and vice versa

- Serial to parallel
 - Use a serial-in, parallel-out shift register

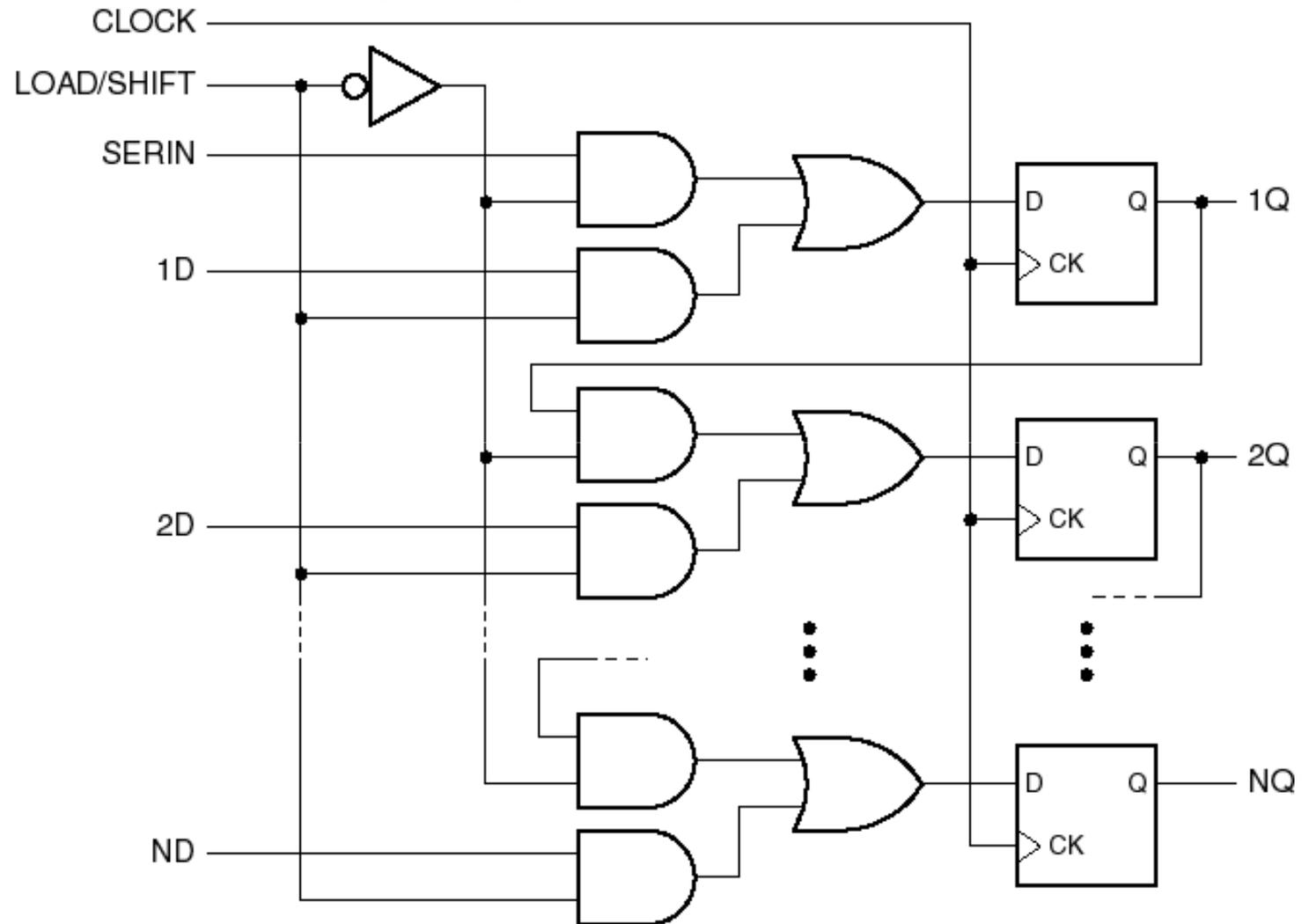


- Parallel to serial
 - Use parallel-in, serial-out shift register



Do both

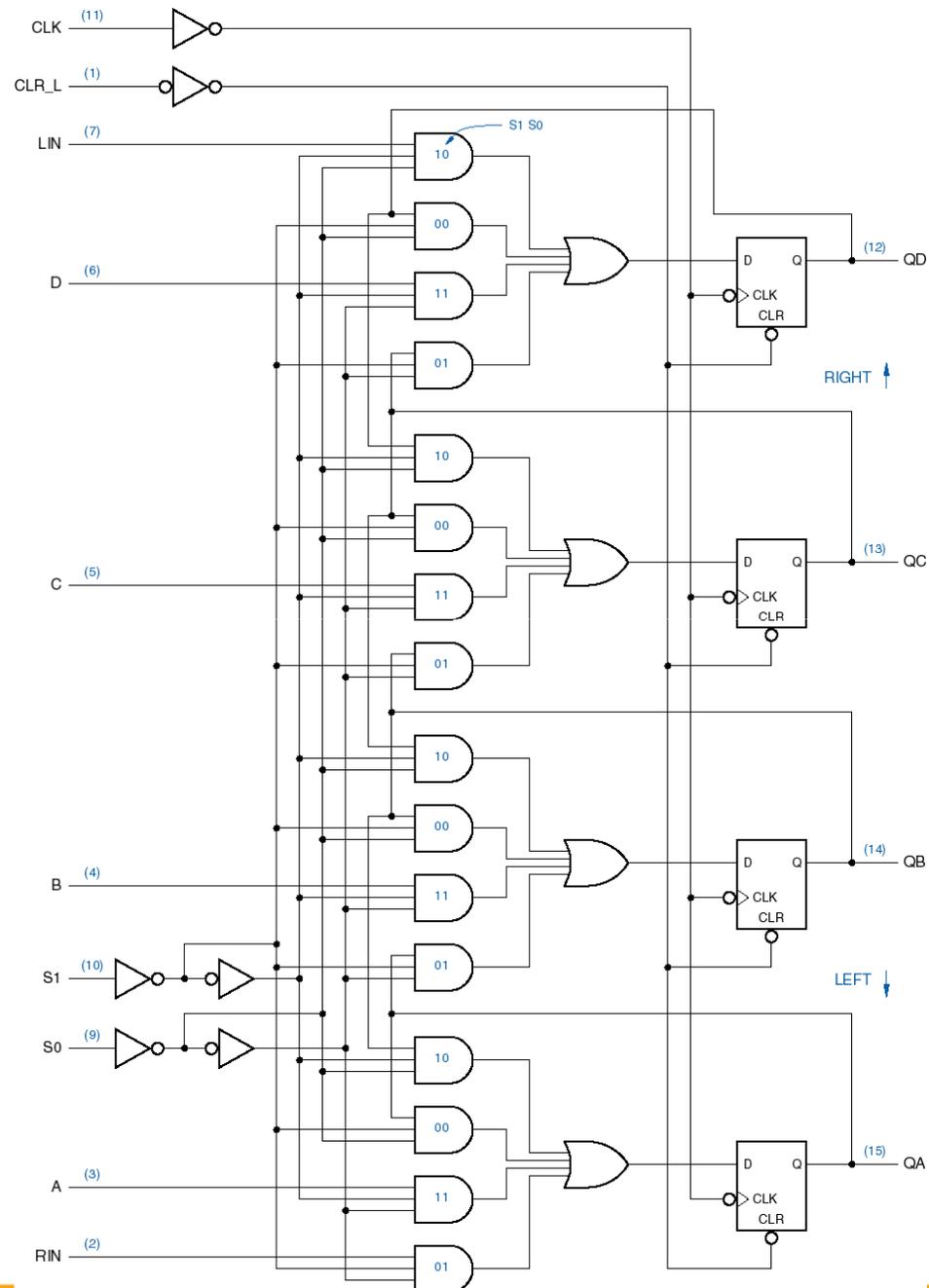
- Parallel-in, parallel-out shift register



“Universal” shift register 74x194

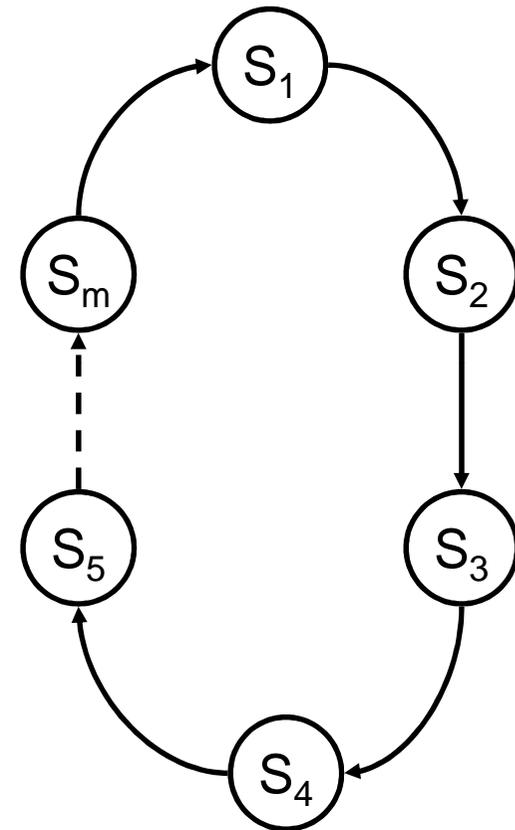
S1	S0	Operation
0	0	Hold
0	1	Shift Up
1	0	Shift Down
1	1	Parallel Load

- S1 & S0 selects which line is connected to D input.
- There's 4 possible inputs to each DFF.



Counters

- A circuit that produces a well-defined output pattern sequence
 - 3 Bit Up-counter: 000, 001, 010, 011, 100, 101, 110, 111, 000, ...
 - 3 Bit Down-counter: 111, 110, 101, 100, 011, 010, 001, 000, 111, ...
 - Binary vs. BCD vs. Gray Code Counters
- The output pattern = **state of the counter**
- Total number of states = **modulus of counter**
 - Counter with m states = **modulus- m counter** or **mod- m counter**
- Counting sequence often shown using a **state diagram** or **state transition diagram**
- A counter is a "degenerate" finite state machine (FSM) circuit where the state is the only output - more on FSM next week

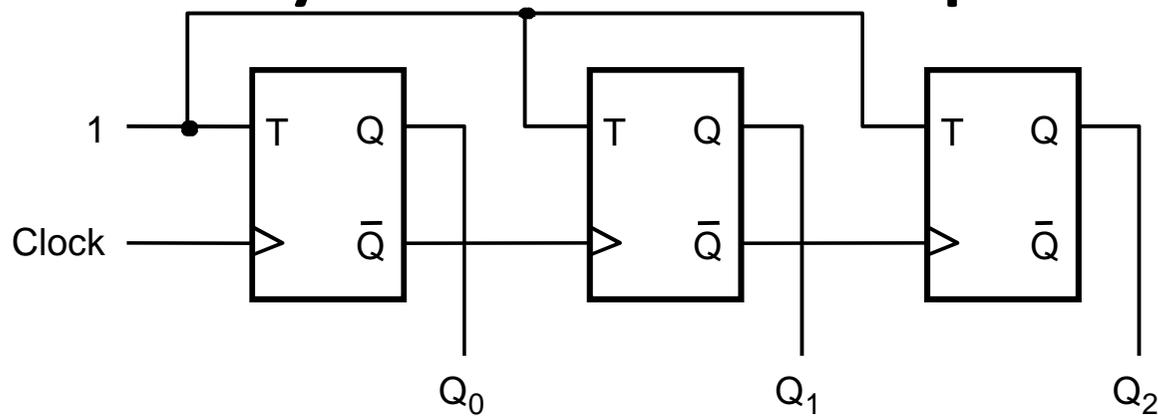


State transition
diagram of a
counter

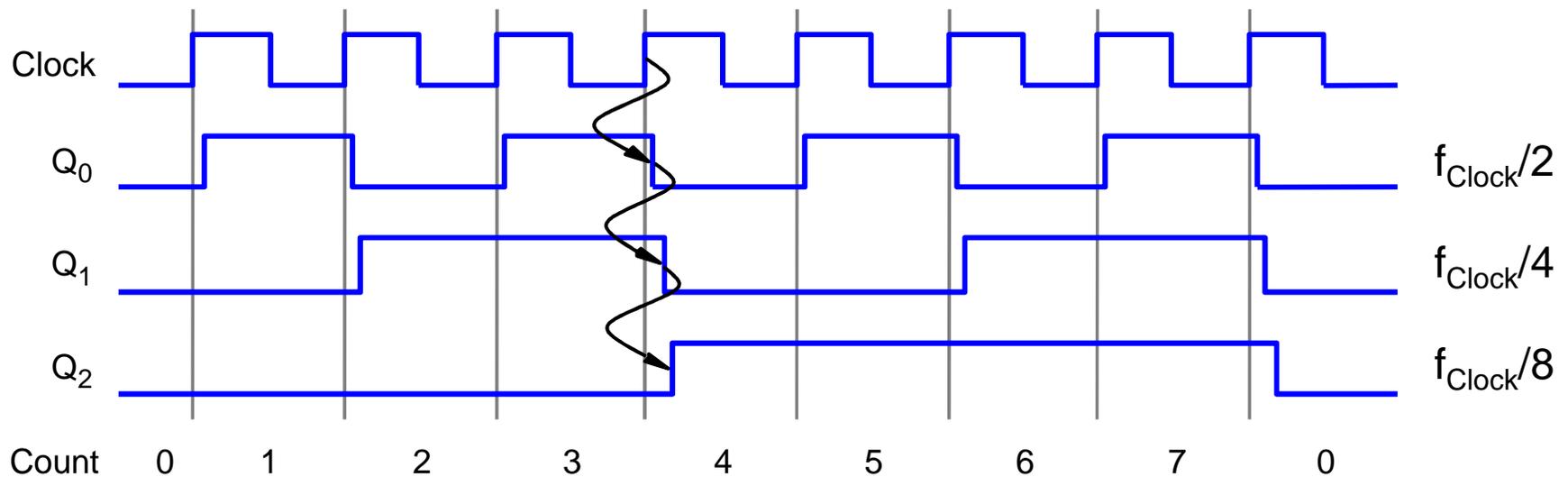
Asynchronous Binary Counters

- **Binary counters** = counters whose counting sequence corresponds to binary numbers
- Modulus of a binary counter is 2^n , where n is # flip-flops
- Also known as **ripple counter** since a change in Q_i flip-flop toggles the Q_{i+1} flip-flop
 - Effect of counting must ripple thru the counter
 - Only first FF connected to clock signal
- Rippling affects overall delay between count pulse and when the count stabilizes
 - Worst case in $n \times t_{pd}$ (t_{pd} is propagation delay of each FF)
- However, ripple counters are useful as frequency dividers
 - Frequency at output of Q_{i+1} flip-flop is half at output of Q_i
 - Frequency of last FF of n -stage counter is $f_{input}/2^n$

A 3-bit Asynchronous Up-Counter



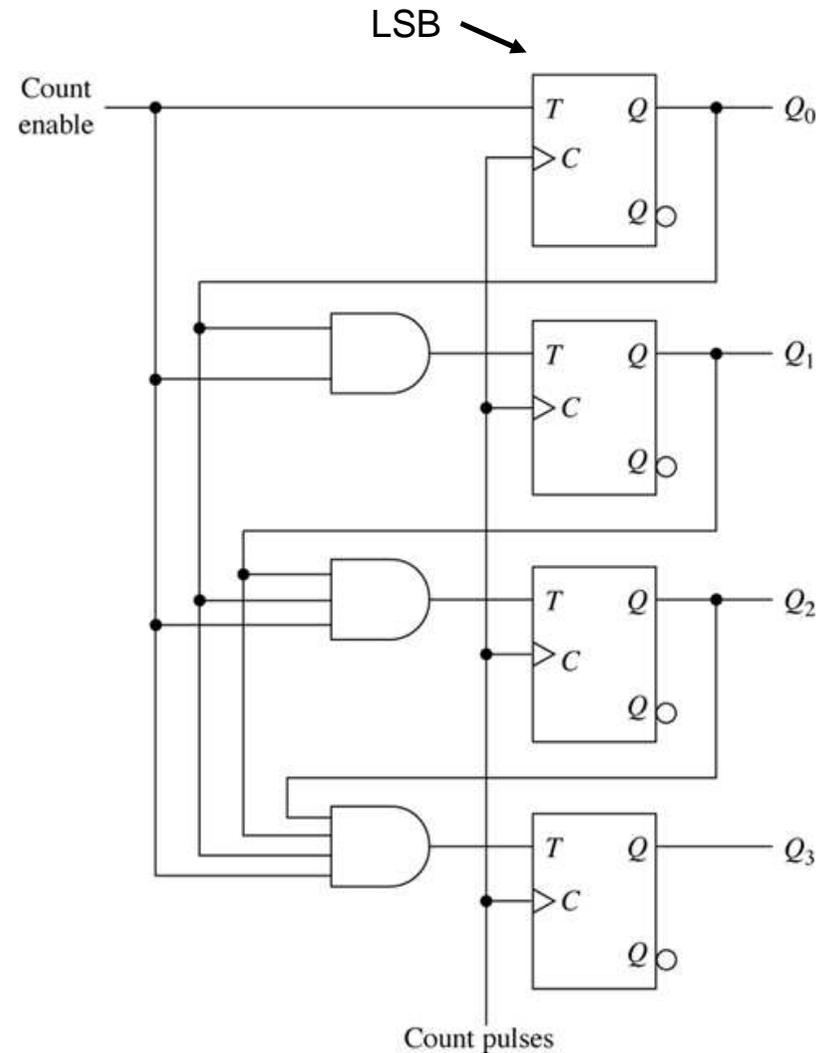
(a) Circuit



(b) Timing diagram showing ripple effect

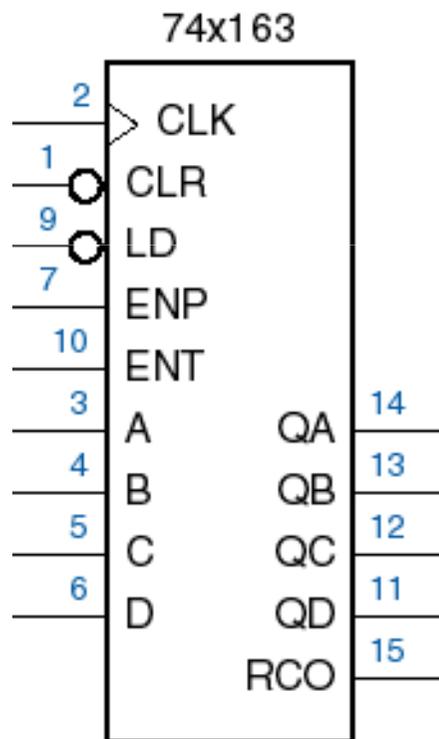
Synchronous Counters

- All FFs are triggered simultaneously (in parallel) by clock input pulses.
- All outputs change simultaneously
- Simple counters use TFF or JKFF
- Only LSB FF has its JK inputs permanently at HIGH level.
- JK inputs of the others FFs are driven by some combination of FF outputs.



74x163

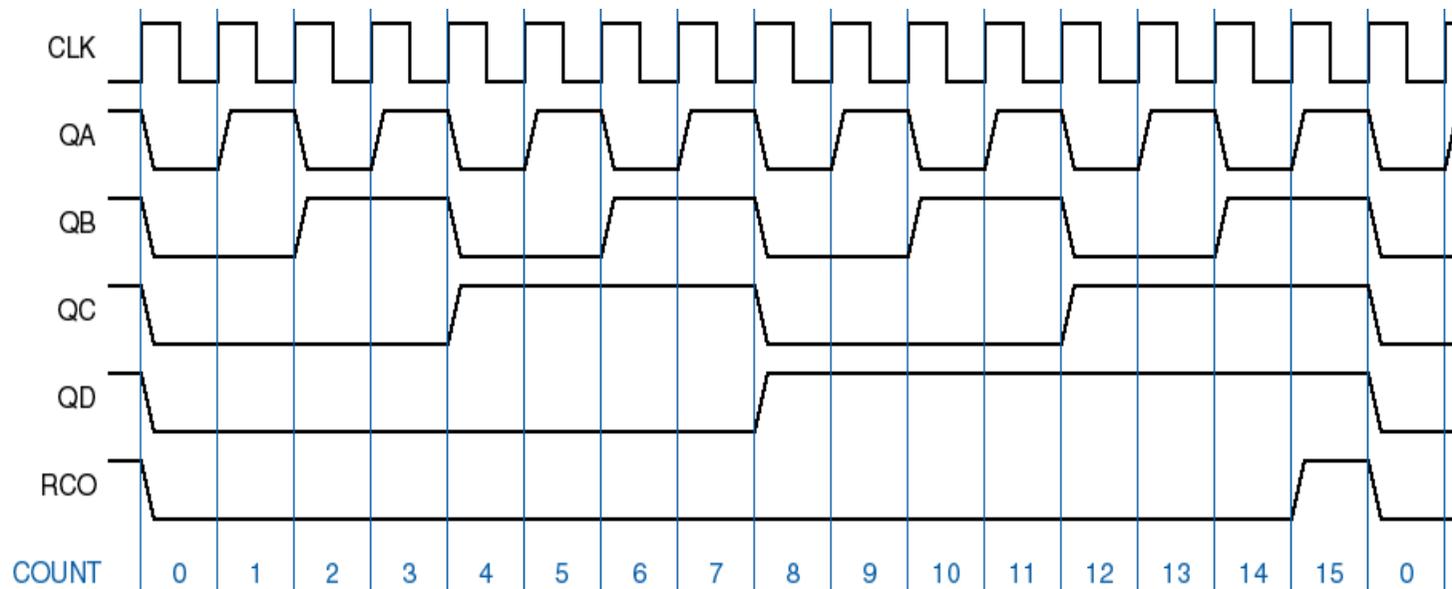
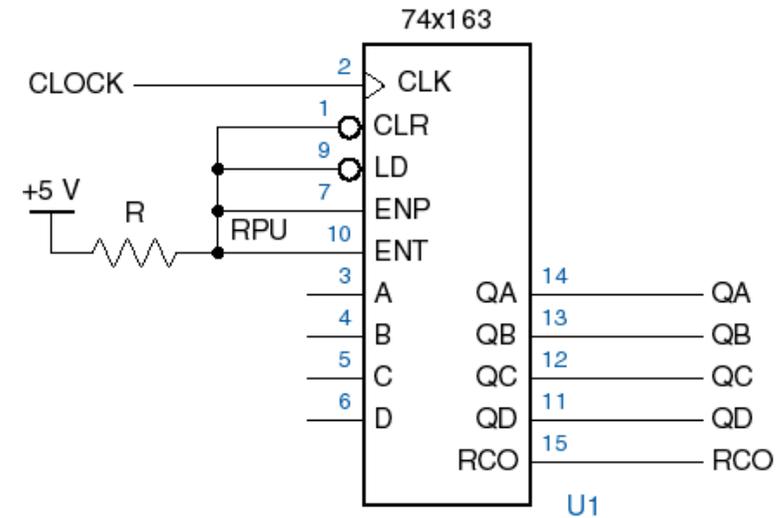
MSI 4-bit counter



Inputs				Current State				Next State			
CLR_L	LD_L	ENT	ENP	QD	QC	QB	QA	QD*	QC*	QB*	QA*
0	x	x	x	x	x	x	x	0	0	0	0
1	0	x	x	x	x	x	x	D	C	B	A
1	1	0	x	x	x	x	x	QD	QC	QB	QA
1	1	x	0	x	x	x	x	QD	QC	QB	QA
1	1	1	1	0	0	0	0	0	0	0	1
1	1	1	1	0	0	0	1	0	0	1	0
1	1	1	1	0	0	1	0	0	0	1	1
1	1	1	1	0	0	1	1	0	1	0	0
1	1	1	1	0	1	0	0	0	1	0	1
1	1	1	1	0	1	0	1	0	1	1	0
1	1	1	1	0	1	1	0	0	1	1	1
1	1	1	1	0	1	1	1	1	0	0	0
1	1	1	1	1	0	0	0	1	0	0	1
1	1	1	1	1	0	0	1	1	0	1	0
1	1	1	1	1	0	1	0	1	0	1	1
1	1	1	1	1	0	1	1	1	1	0	0
1	1	1	1	1	1	1	0	0	1	1	0
1	1	1	1	1	1	1	1	0	1	1	1
1	1	1	1	1	1	1	1	0	0	0	0

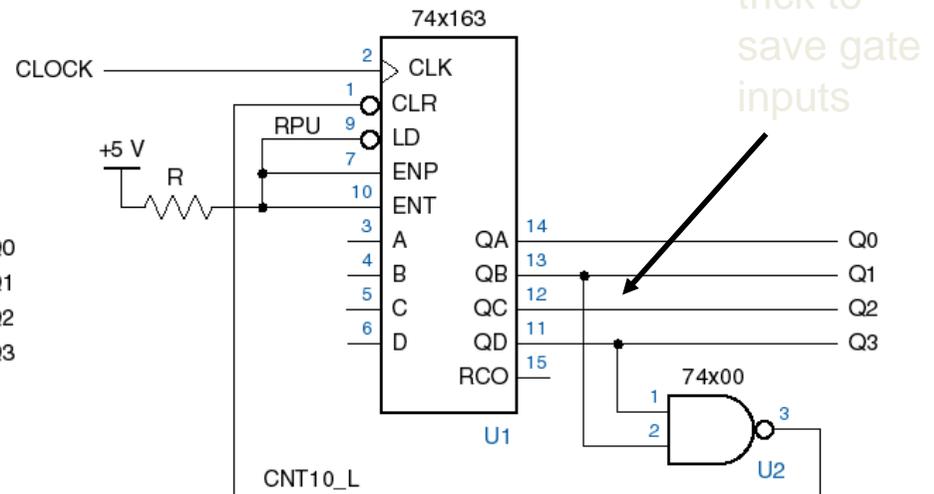
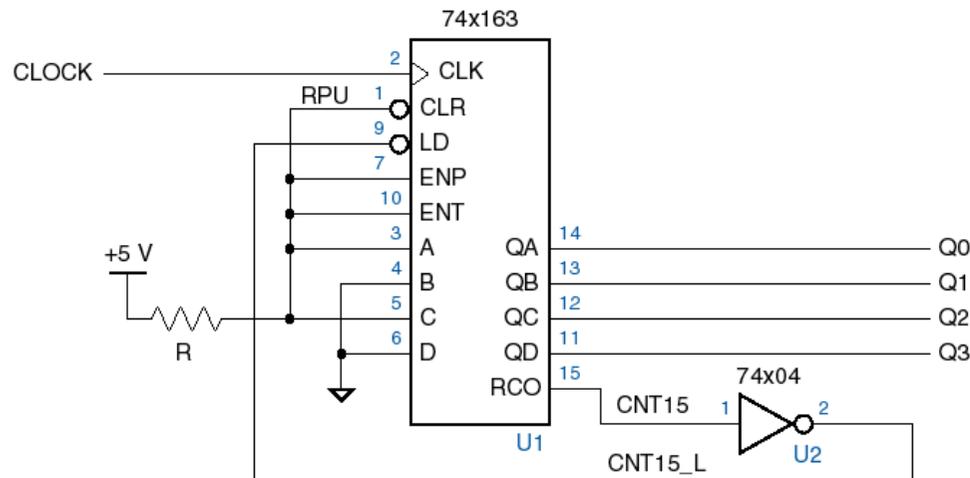
Free-Running 4-bit '163 Counter

- “divide-by-16” counter

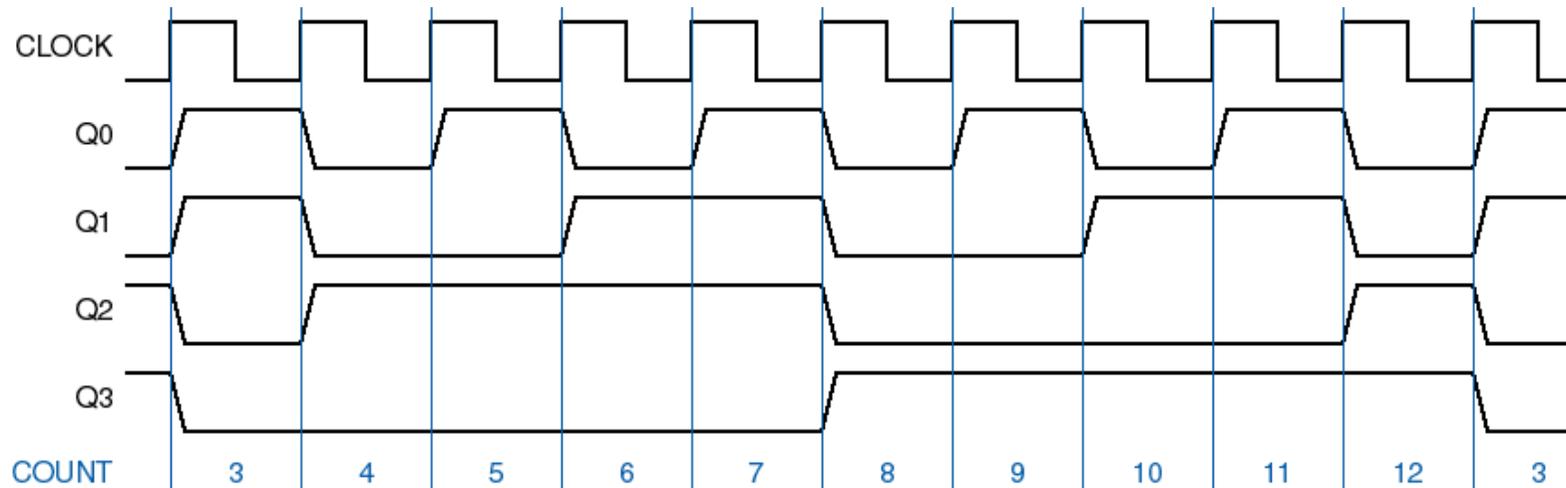
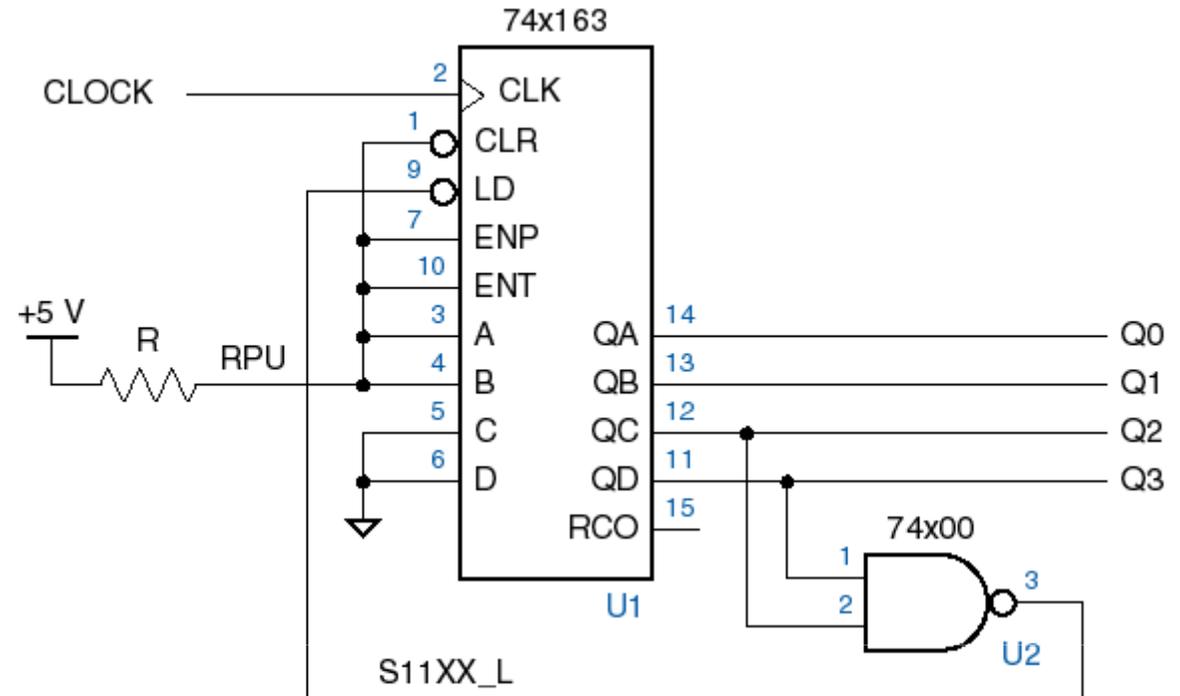


Modified Counting sequence: mod-11 Counter

- Load 0101 (5) after Count = 15
- 5, 6, 7, 8, 9, 10, 11, 12, 13, 14, 15, 5, 6, ...
- Clear after Count = 1010 (10)
- 0, 1, 2, 3, 4, 5, 6, 7, 8, 9, 10, 0, 1, 2, 3, ...

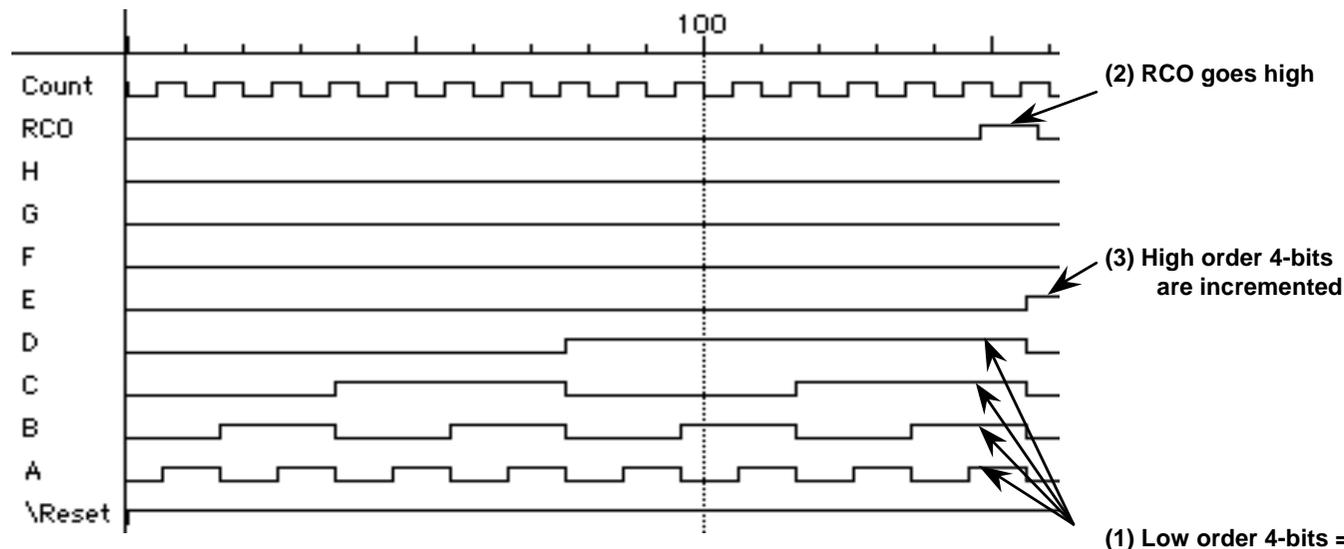
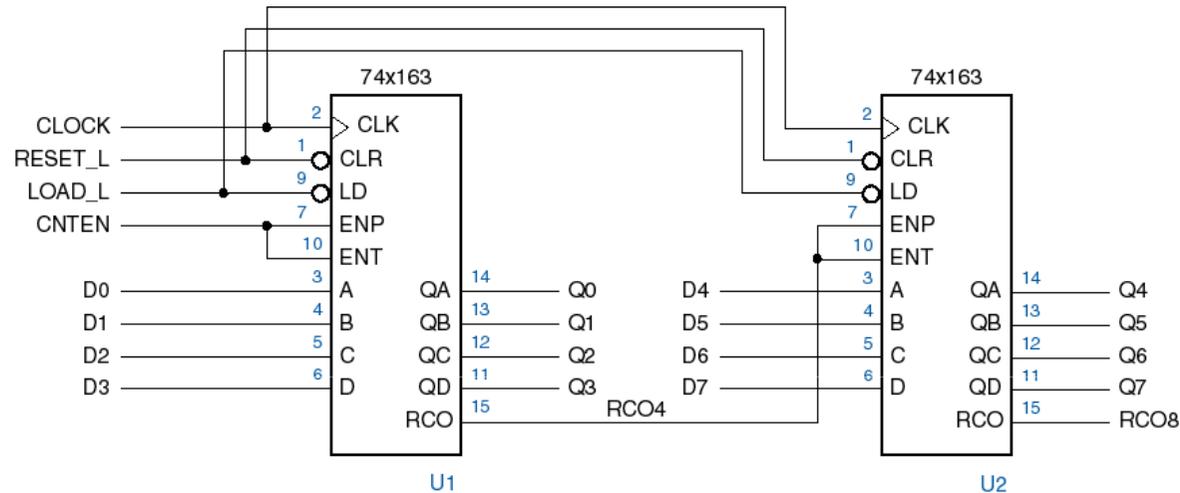


Counting from 3 to 12



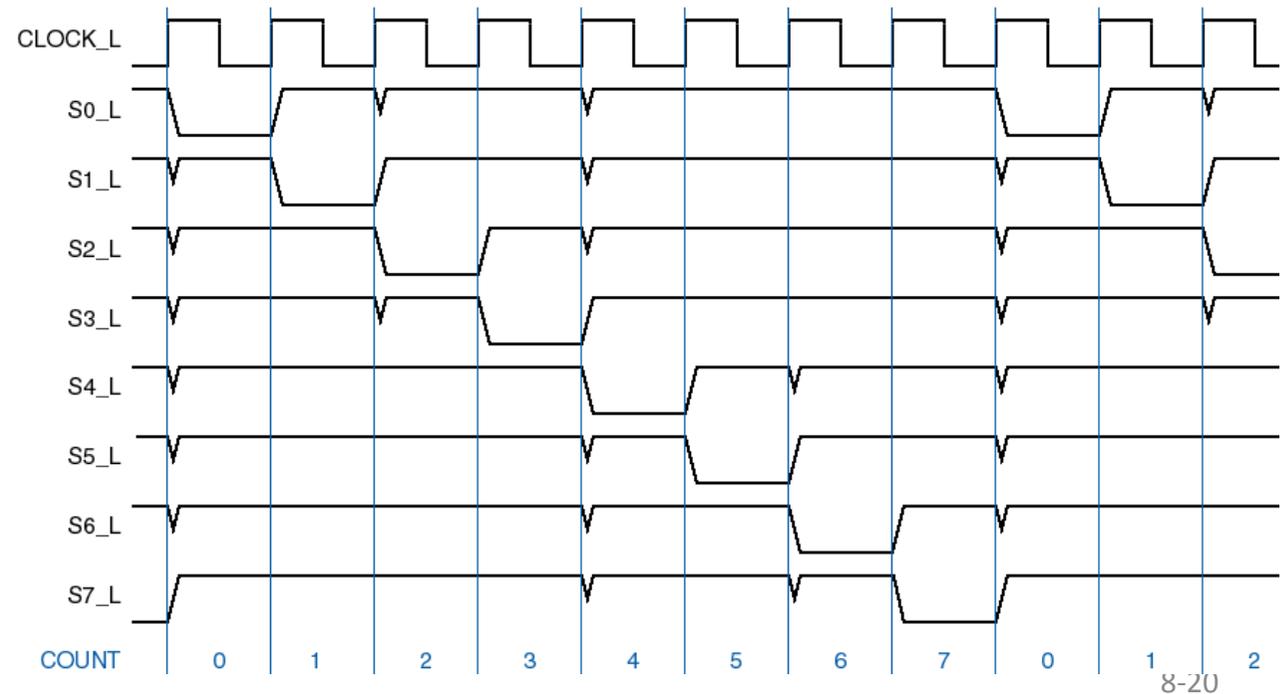
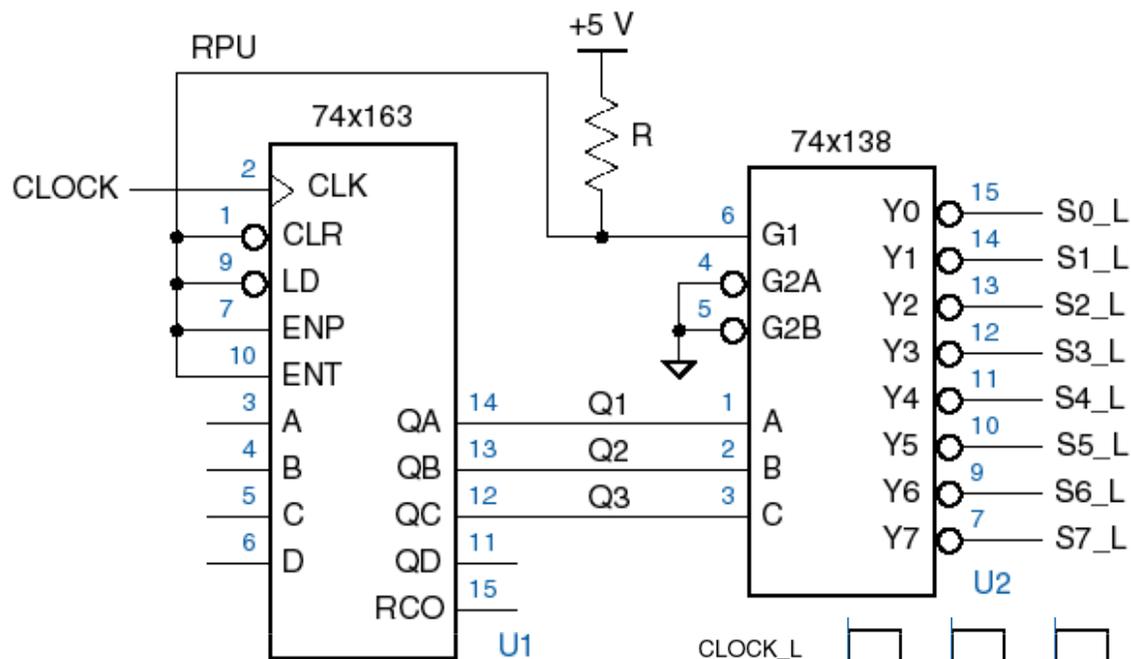
Cascading Counters

- RCO (ripple carry out) is asserted in state 15, if ENT is asserted.



- First stage RCO enables second stage for counting
- RCO asserted soon after stage enters state 1111
- also a function of the T Enable
- Downstream stages lag in their 1111 to 0000 transitions
- Affects Count period and decoding logic

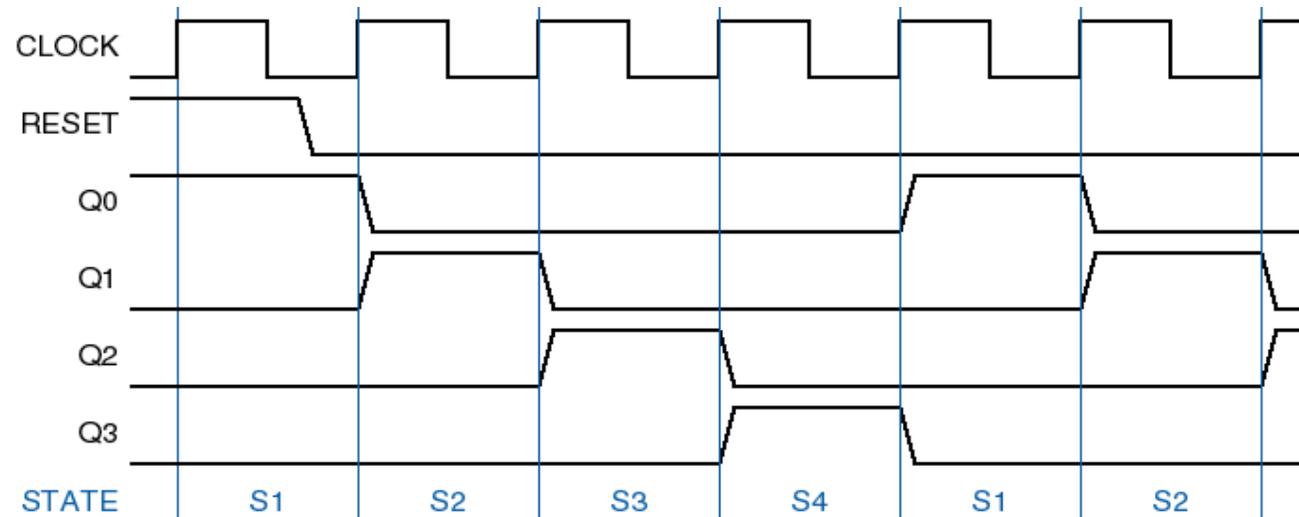
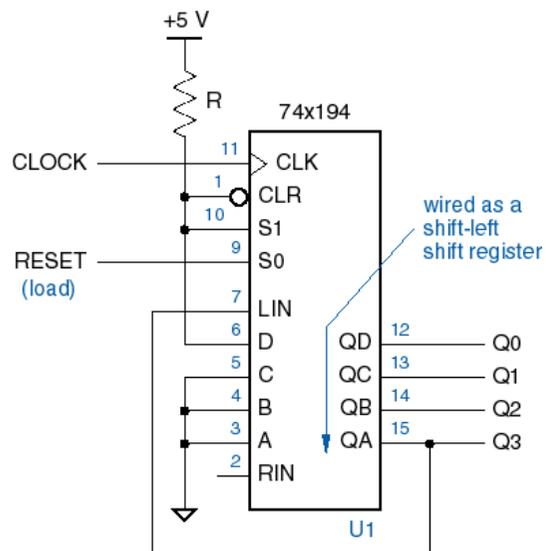
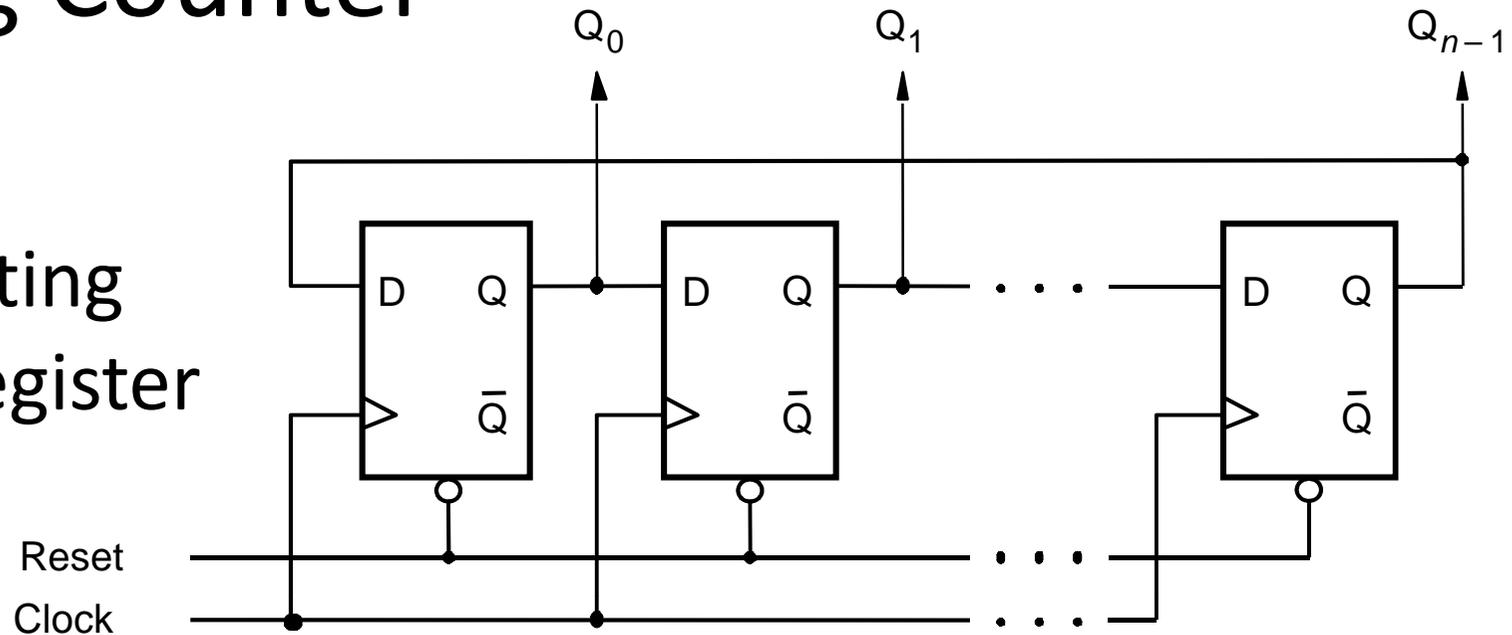
Decoding binary- counter states



- Glitches may or may not be a concern.

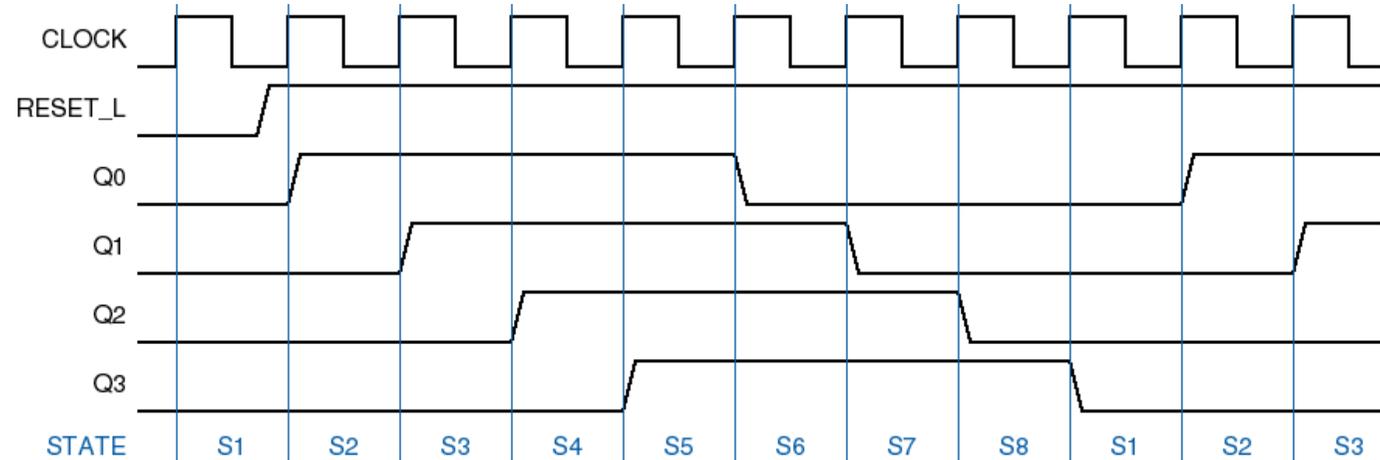
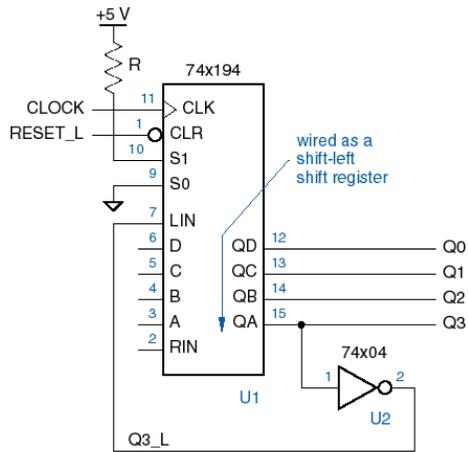
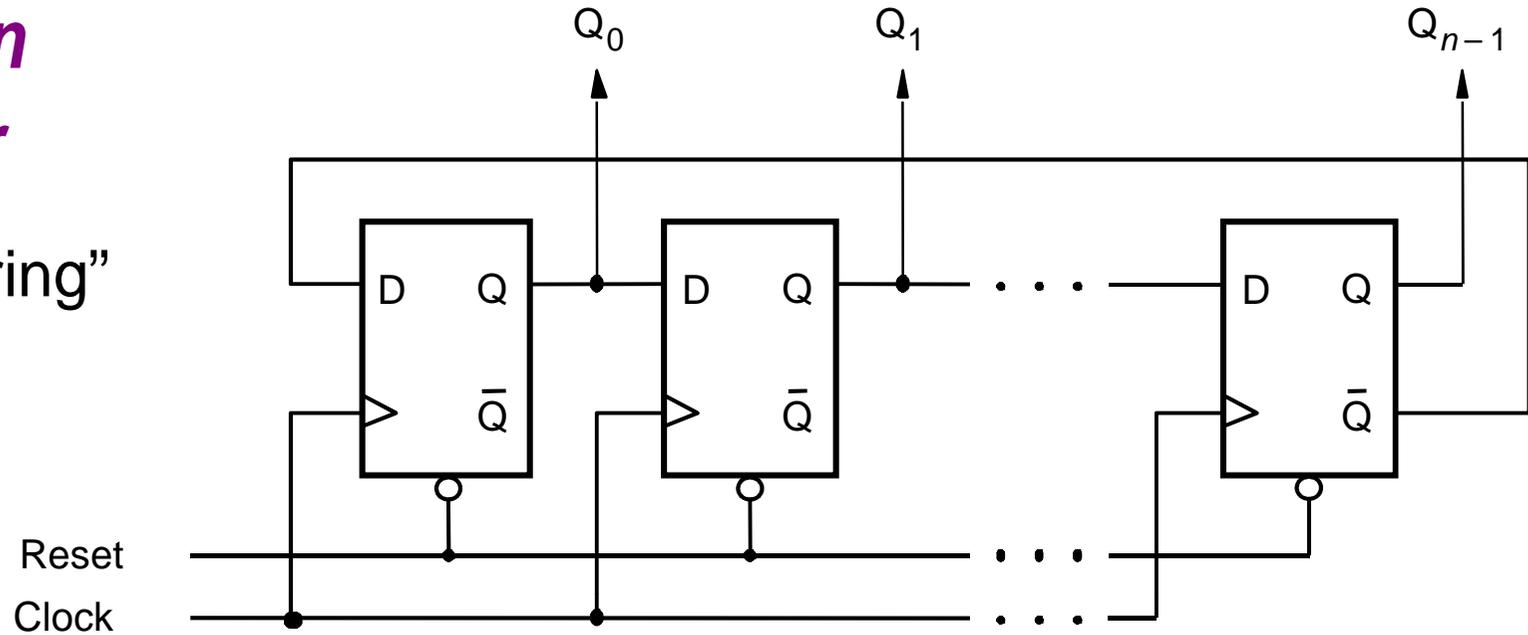
Ring Counter

- Is a circulating shift register



Johnson Counter

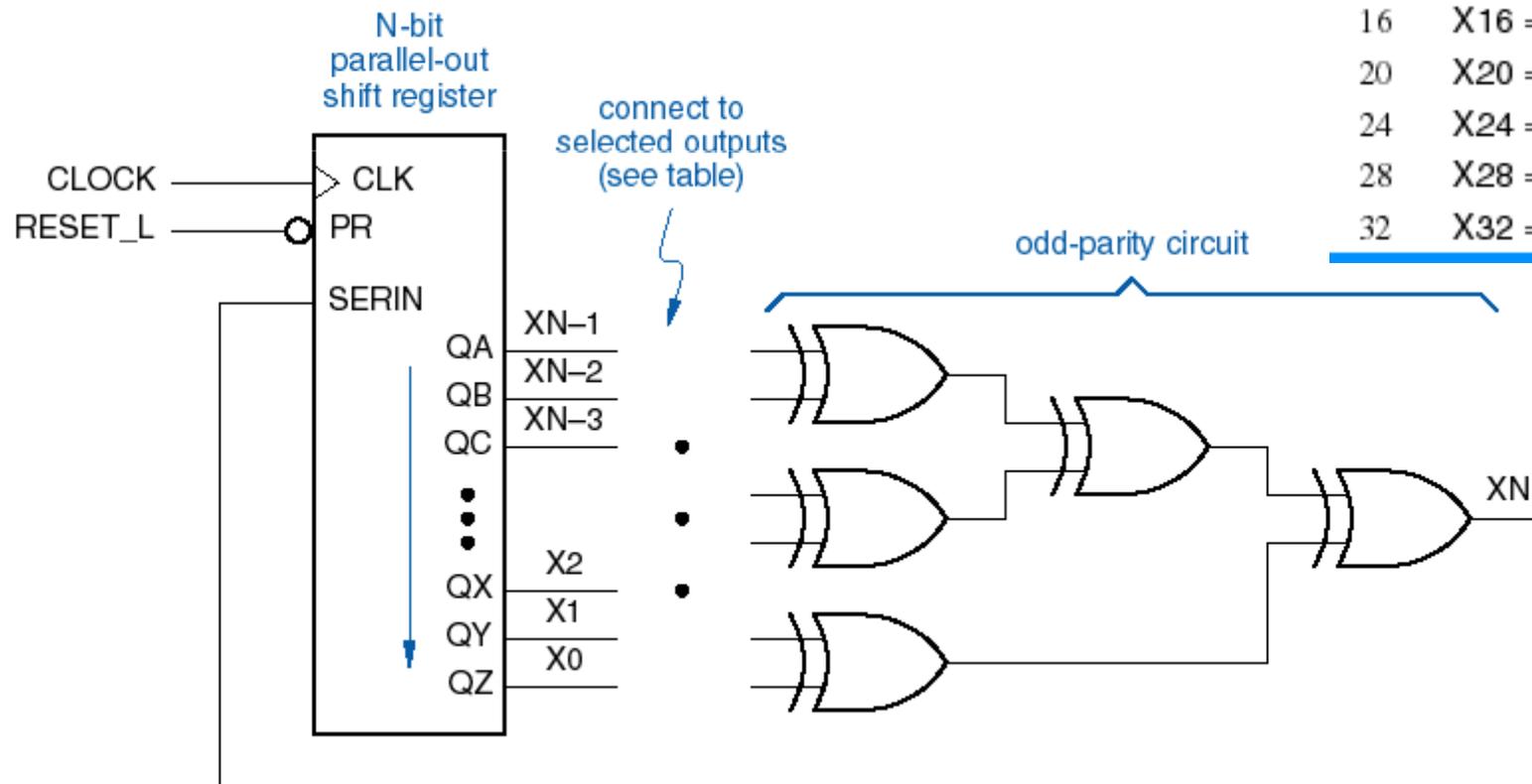
- “Twisted ring” counter



LFSR Counters

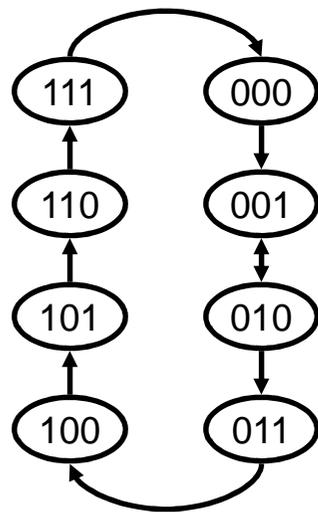
- Pseudo-random number generator
- $2^n - 1$ states before repeating
- Same circuits used in CRC error checking in Ethernet networks, etc.

n	Feedback Equation
2	$X_2 = X_1 \oplus X_0$
3	$X_3 = X_1 \oplus X_0$
4	$X_4 = X_1 \oplus X_0$
5	$X_5 = X_2 \oplus X_0$
6	$X_6 = X_1 \oplus X_0$
7	$X_7 = X_3 \oplus X_0$
8	$X_8 = X_4 \oplus X_3 \oplus X_2 \oplus X_0$
12	$X_{12} = X_6 \oplus X_4 \oplus X_1 \oplus X_0$
16	$X_{16} = X_5 \oplus X_4 \oplus X_3 \oplus X_0$
20	$X_{20} = X_3 \oplus X_0$
24	$X_{24} = X_7 \oplus X_2 \oplus X_1 \oplus X_0$
28	$X_{28} = X_3 \oplus X_0$
32	$X_{32} = X_{22} \oplus X_2 \oplus X_1 \oplus X_0$



Design of 3-bit Binary Upcounter

- This procedure can be generalized to implement ANY finite state machine
- Counters are a very simple way to start:
 - no decisions on what state to advance to next
 - current state is the output



State Transition Diagram for
3-bit binary upcounter

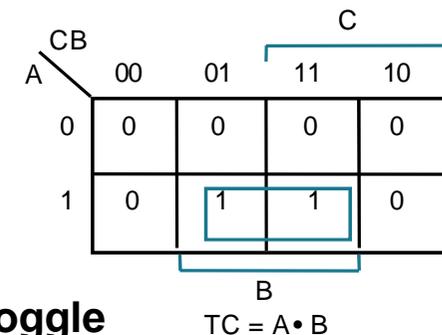
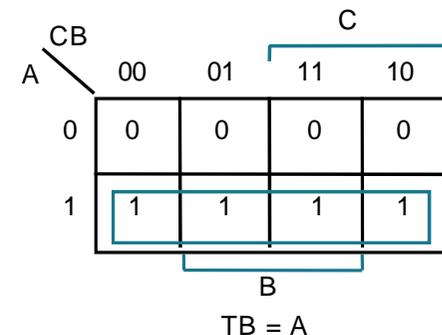
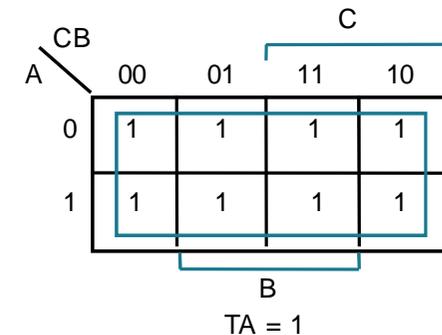
Present State			Next State		
C	B	A	C+	B+	A+
0	0	0	0	0	1
0	0	1	0	1	0
0	1	0	0	1	1
0	1	1	1	0	0
1	0	0	1	0	1
1	0	1	1	1	0
1	1	0	1	1	1
1	1	1	0	0	0

State Transition Table

Design of 3-bit Binary Upcounter

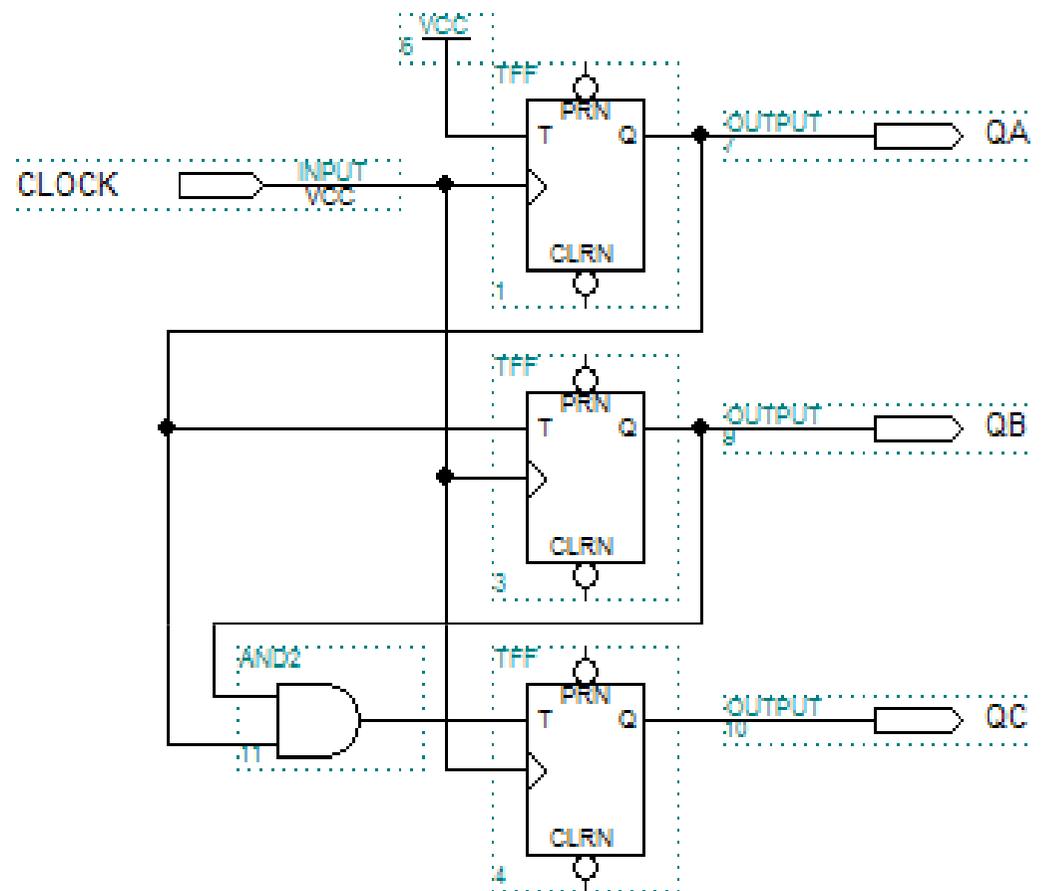
- Let's implement with Toggle Flipflops
- What inputs must be presented to the T FFs to get them to change to the desired state bit?
- This is called "Remapping the Next State Function"

Present State			Next State			Flipflop Inputs		
C	B	A	C+	B+	A+	TC	TB	TA
0	0	0	0	0	1	0	0	1
0	0	1	0	1	0	0	1	1
0	1	0	0	1	1	0	0	1
0	1	1	1	0	0	1	1	1
1	0	0	1	0	1	0	0	1
1	0	1	1	1	0	0	1	1
1	1	0	1	1	1	0	0	1
1	1	1	0	0	0	1	1	1

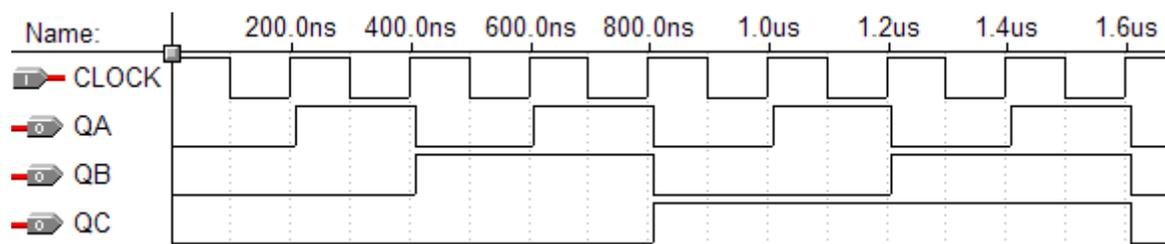


K-maps for Toggle Inputs:

Design of 3-bit Binary Upcounter



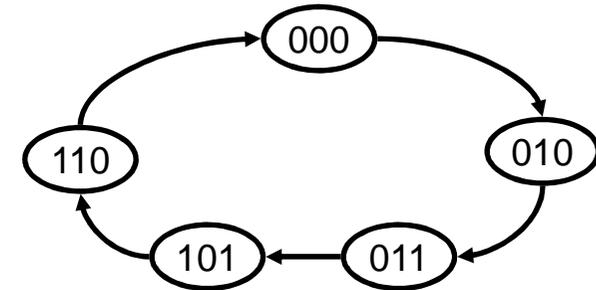
Timing Diagram:



Design of Counter with Complex Count Sequence

Step 1: Derive the State Transition Diagram

Count sequence: 000, 010, 011, 101, 110



Present State			Next State		
C	B	A	C+	B+	A+
0	0	0	0	1	0
0	0	1	X	X	X
0	1	0	0	1	1
0	1	1	1	0	1
1	0	0	X	X	X
1	0	1	1	1	0
1	1	0	0	0	0
1	1	1	X	X	X

Step 2: State Transition Table

Note the Don't Care conditions

Design of Counter with Complex Count Sequence

Step 3: Choose Flipflop Type for Implementation
Use **Excitation Table** to Remap Next State Functions

Q	Q+	T
0	0	0
0	1	1
1	0	1
1	1	0

Present State			Next State			Toggle Inputs		
C	B	A	C+	B+	A+	TC	TB	TA
0	0	0	0	1	0	0	1	0
0	0	1	X	X	X	X	X	X
0	1	0	0	1	1	0	0	1
0	1	1	1	0	1	1	1	0
1	0	0	X	X	X	X	X	X
1	0	1	1	1	0	0	1	1
1	1	0	0	0	0	1	1	0
1	1	1	X	X	X	X	X	X

		C				
		00	01	11	10	
A	CB					
0	0	0	0	1	X	TC
1	1	X	1	X	0	

		C				
		00	01	11	10	
A	CB					
0	0	1	0	1	X	TB
1	1	X	1	X	1	

		C				
		00	01	11	10	
A	CB					
0	0	0	1	0	X	TA
1	1	X	0	X	1	

$$TC = A' C + A C' = A \text{ xor } C$$

$$TB = A + B' + C$$

$$TA = A' B C' + B' C'$$

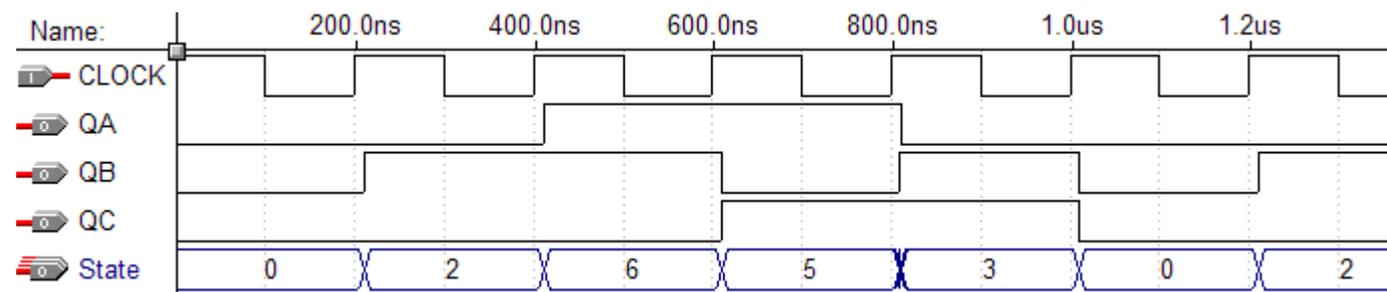
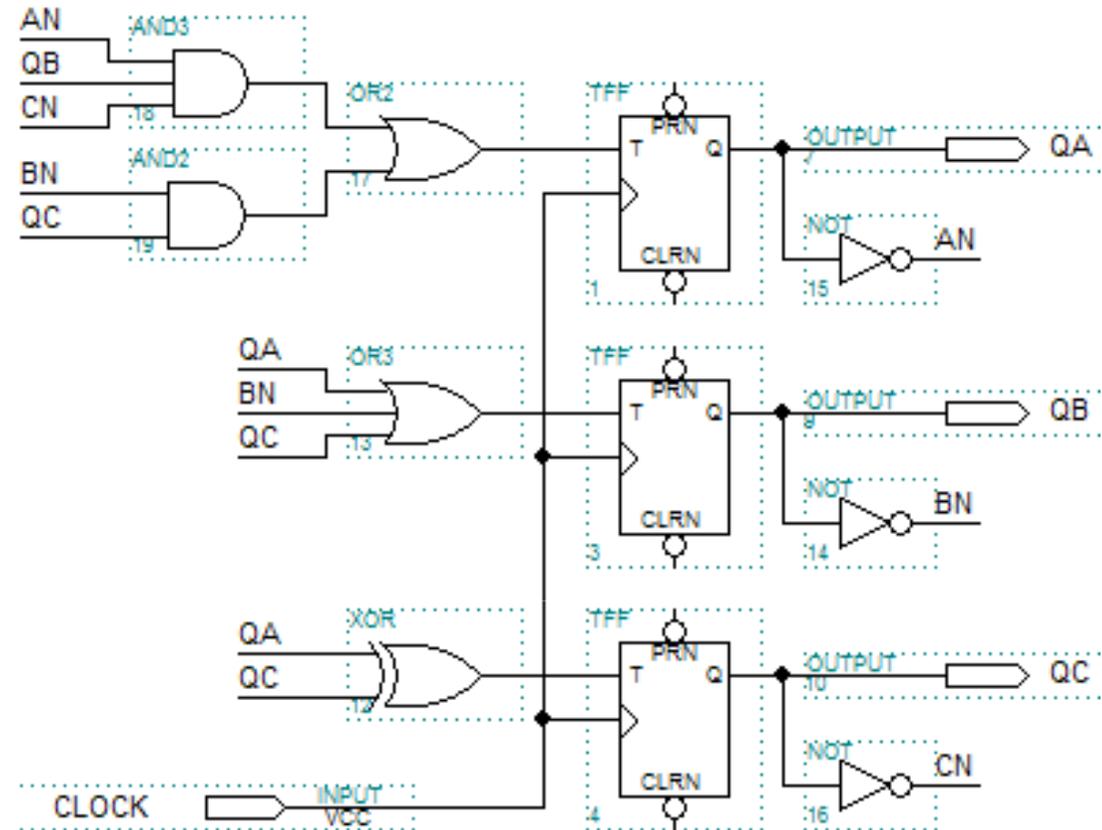
Remapped Next
State Functions

Counter Design Procedure

Resulting Logic:

5 Gates
13 Input Literals +
Flipflop connections

Timing Waveform:



Implementation with Different Kinds of FFs

SR Flipflops

Continuing with the 000, 010, 011, 101, 110, 000, ... counter example

SR Excitation Table
 $Q^+ = S + R'Q$

Q	Q ⁺	S	R
0	0	0	X
0	1	1	0
1	0	0	1
1	1	X	0

Present State			Next State			Remapped Next State					
C	B	A	C ⁺	B ⁺	A ⁺	SC	RC	SB	RB	SA	RA
0	0	0	0	1	0	0	X	1	0	0	X
0	0	1	X	X	X	X	X	X	X	X	X
0	1	0	0	1	1	0	X	X	0	1	0
0	1	1	1	0	1	1	0	0	1	X	0
1	0	0	X	X	X	X	X	X	X	X	X
1	0	1	1	1	0	X	0	1	0	0	1
1	1	0	0	0	0	0	1	0	1	0	X
1	1	1	X	X	X	X	X	X	X	X	X

Implementation with Different Kinds of FFs

SR FFs Continued

A \ CB		C				
		00	01	11	10	
A	0	X	X	1	X	<i>RC</i>
	1	X	0	X	0	
		B				

A \ CB		C				
		00	01	11	10	
A	0	0	0	0	X	<i>SC</i>
	1	X	1	X	X	
		B				

A \ CB		C				
		00	01	11	10	
A	0	0	0	1	X	<i>RB</i>
	1	X	1	X	0	
		B				

A \ CB		C				
		00	01	11	10	
A	0	1	X	0	X	<i>SB</i>
	1	X	0	X	1	
		B				

A \ CB		C				
		00	01	11	10	
A	0	X	0	X	X	<i>RA</i>
	1	X	0	X	1	
		B				

A \ CB		C				
		00	01	11	10	
A	0	0	1	0	X	<i>SA</i>
	1	X	X	X	0	
		B				

$$RC = A'$$

$$SC = A$$

$$RB = AB + BC = B(A+C)$$

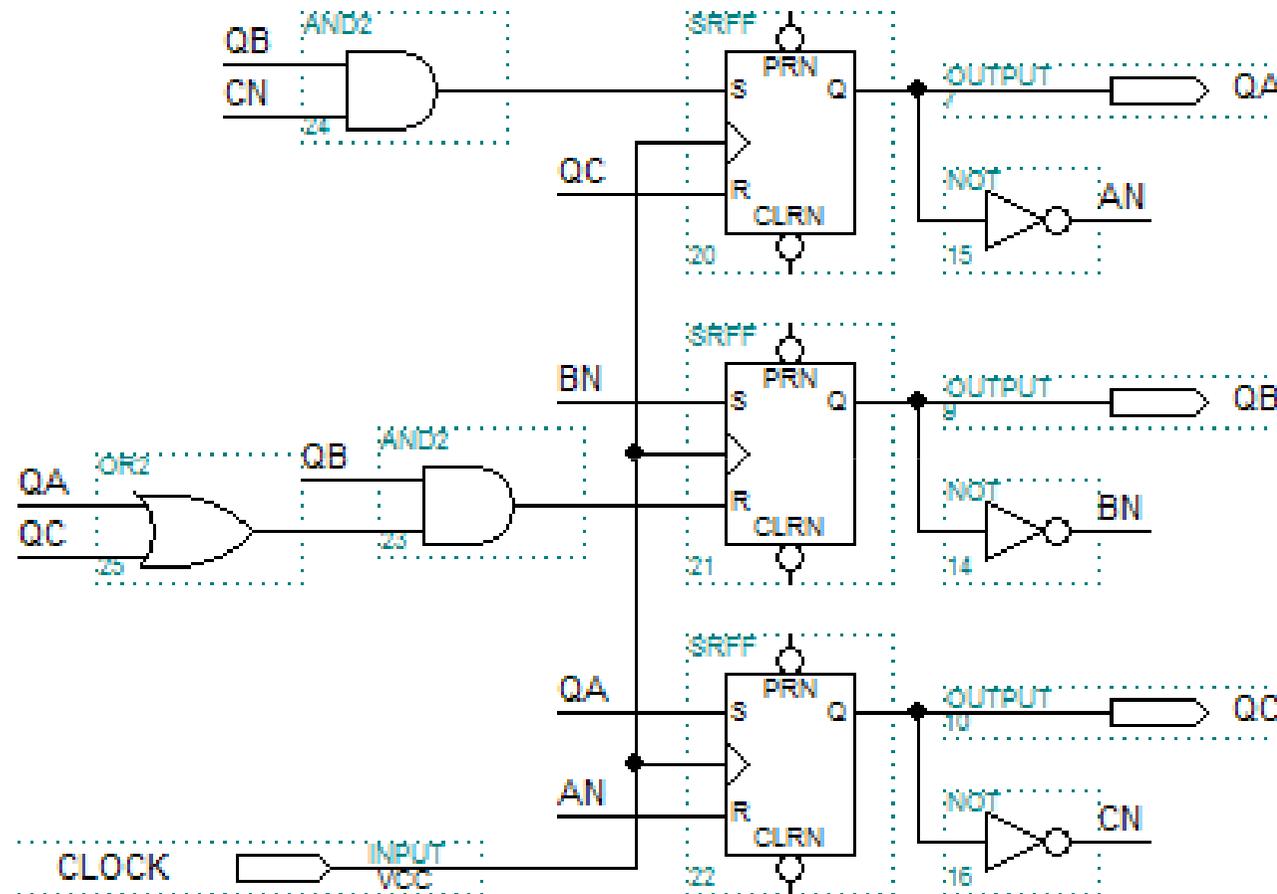
$$SB = B'$$

$$RA = C$$

$$SA = BC'$$

Implementation With Different Kinds of FFs

SR FFs
Continued



Resulting Logic Level Implementation:
3 Gates, 11 Input Literals + Flipflop connections

Implementation with Different FF Types

JK FFs

JK Excitation Table
 $Q_+ = JQ' + K'Q$

Q	Q ₊	J	K
0	0	0	X
0	1	1	X
1	0	X	1
1	1	X	0

Present State			Next State			Remapped Next State					
C	B	A	C ₊	B ₊	A ₊	JC	KC	JB	KB	JA	KA
0	0	0	0	1	0	0	X	1	X	0	X
0	0	1	X	X	X	X	X	X	X	X	X
0	1	0	0	1	1	0	X	X	0	1	X
0	1	1	1	0	1	1	X	X	1	X	0
1	0	0	X	X	X	X	X	X	X	X	X
1	0	1	1	1	0	X	0	1	X	X	1
1	1	0	0	0	0	X	1	X	1	0	X
1	1	1	X	X	X	X	X	X	X	X	X

Implementation with Different FF Types

JK FFs Continued

		C				
		00	01	11	10	
A	0	0	0	X	X	<i>JC</i>
	1	X	1	X	X	
		B				

		C				
		00	01	11	10	
A	0	X	X	1	X	<i>KC</i>
	1	X	X	X	0	
		B				

$$JC = A$$

$$KC = A'$$

		C				
		00	01	11	10	
A	0	1	X	X	X	<i>JB</i>
	1	X	X	X	1	
		B				

		C				
		00	01	11	10	
A	0	X	0	1	X	<i>KB</i>
	1	X	1	X	X	
		B				

$$JB = 1$$

$$KB = A + C$$

		C				
		00	01	11	10	
A	0	0	1	0	X	<i>JA</i>
	1	X	X	X	X	
		B				

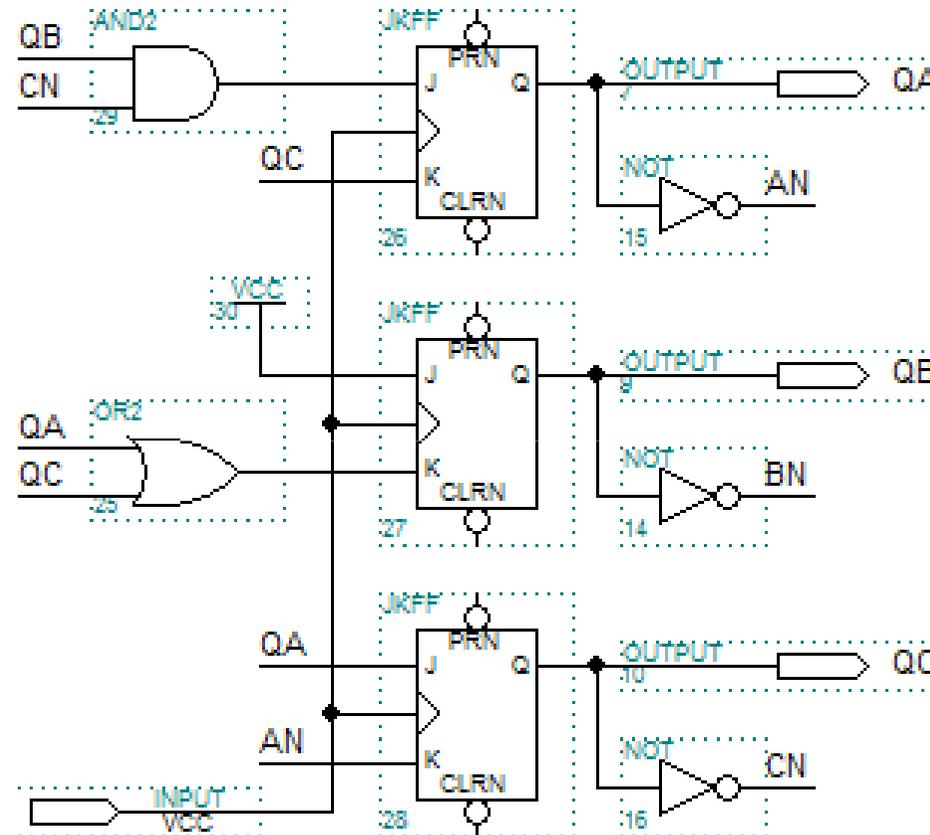
		C				
		00	01	11	10	
A	0	X	X	X	X	<i>KA</i>
	1	X	0	X	1	
		B				

$$JA = B C'$$

$$KA = C$$

Implementation with Different FF Types

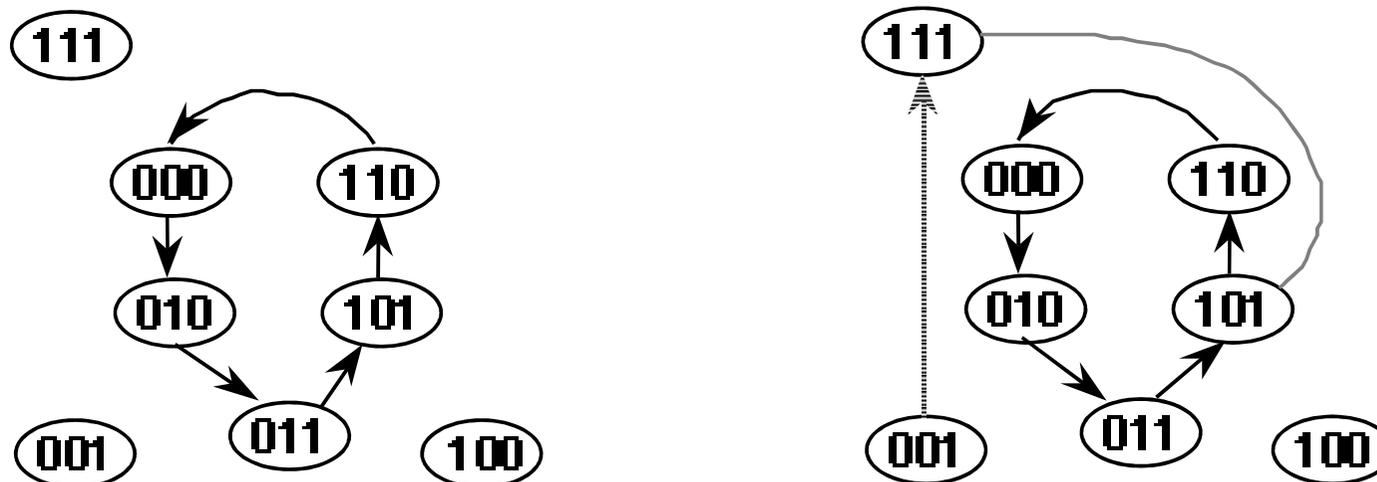
JK FFs Continued



Resulting Logic Level Implementation:
2 Gates, 10 Input Literals + Flipflop Connections

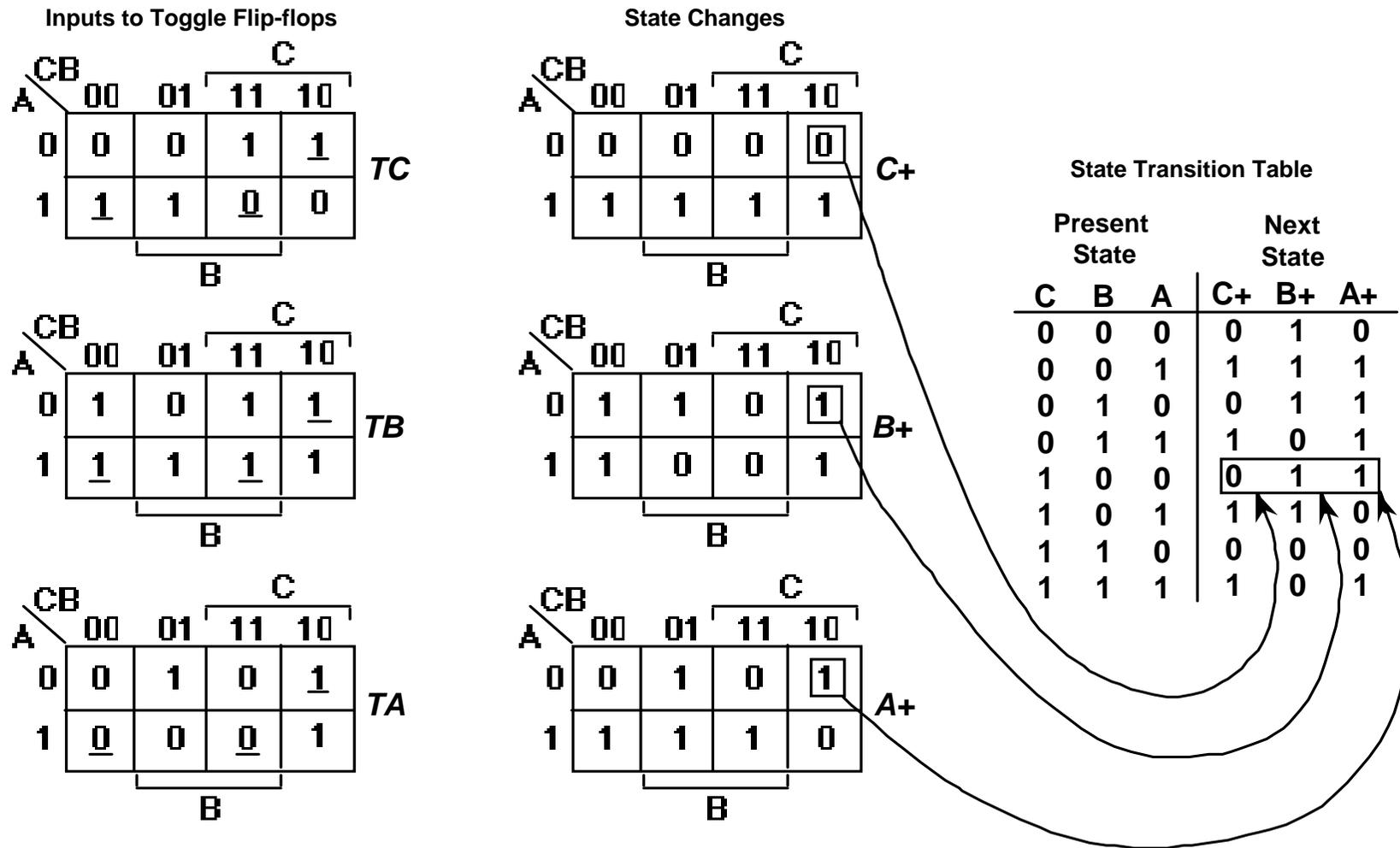
Avoiding Ambiguous States

- Problem with counter with modulo $< 2^n$
 - At power-up, counter may be in ANY possible state
 - Designer must guarantee that it (eventually) enters a valid state
 - Especially a problem for counters that validly use a subset of states
- Self-Starting Counters
 - Design counter so that even invalid states eventually transition to valid state



*Two Self-Starting State Transition Diagrams
for the Example Counter*

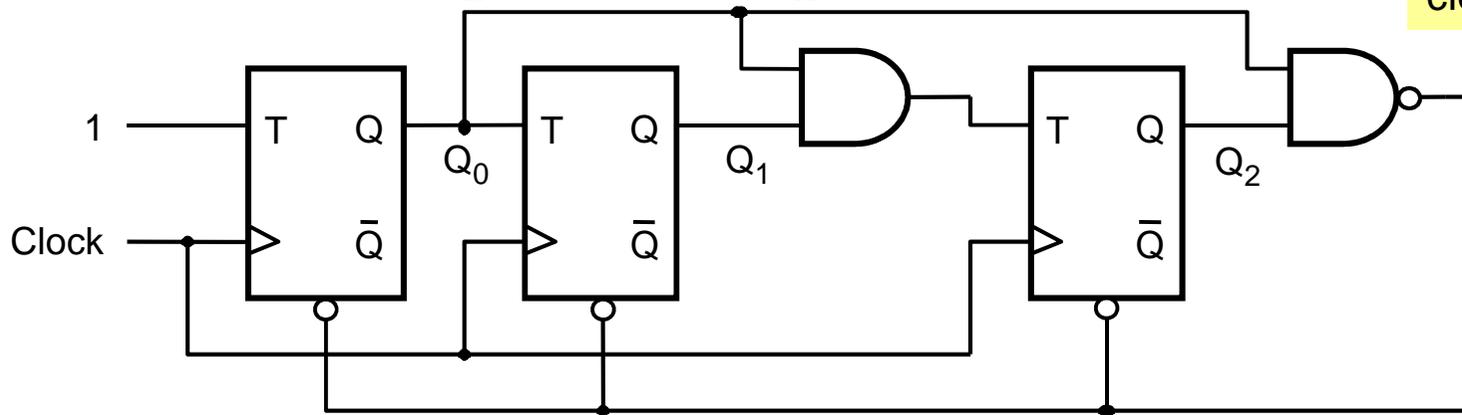
Self-Starting Counters



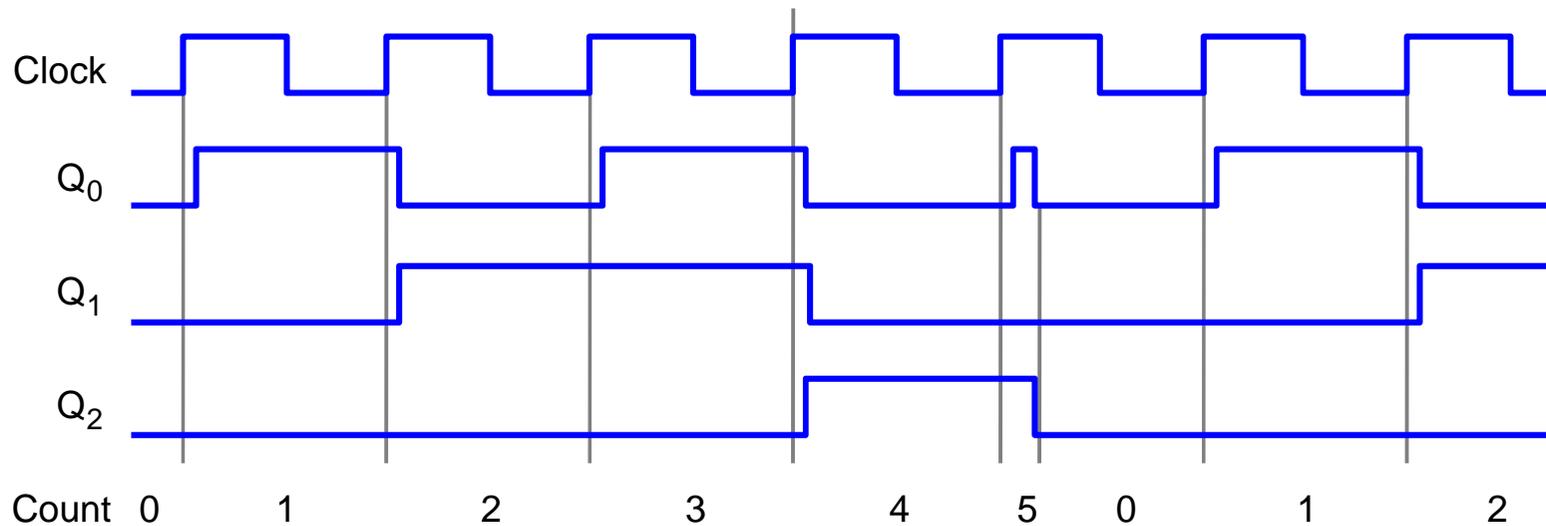
Deriving State Transition Table from Don't Care Assignment

Avoiding Ambiguous States

Never use
counters with
asynchronous
clear



(a) A modulo-6 counter with asynchronous clear



(b) Timing diagram

Counter Implementation with Different FF Types

- T FFs well suited for straightforward binary counters
 - But yielded worst gate and literal count for this example (coz it's not straightforward !)
- No reason to choose SR over JK FFs: it is a proper subset of JK
 - SR FFs don't really exist anyway
- JK FFs yielded lowest gate count
 - Tend to yield best choice for **packaged logic** where gate count is key
- D FFs yield simplest design procedure
 - Best literal count
 - D storage devices very transistor efficient in VLSI
 - Other flipflops most likely implemented using DFF in VLSI/FPGA
 - Best choice where area/literal count is the key